

# Perfect PSK31 Signals, Automatically.

*Building a digital field strength meter for your PSK31 station*

by

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No need to take notes! This entire presentation is available at [www.ssiserver.com/info/pskmeter](http://www.ssiserver.com/info/pskmeter)



# What we will cover this hour...

- PSK31 signalling (how it works)
- A PSK31 station and its components
- Tuning up in PSK31
- Over modulation: the audio level problem
- Building an RF monitor to solve the problem
- Circuit diagram and hardware details
- Firmware and Software routines
- Performance
- Where we go from here

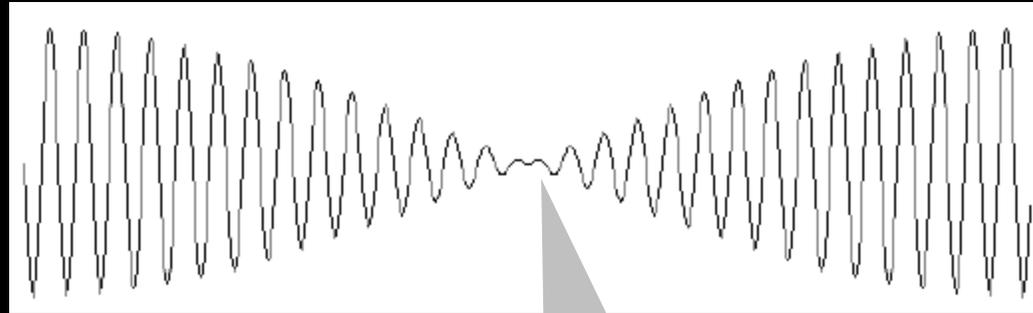
# Digital Modes (character-based data):

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- ✓ CW
- ✓ RTTY
- ✓ Packet
- ✓ APRS
- ✓ SSTV
- ✓ *PSK-31 (and now PSK-63, too)*

# Phase shift keying—the basics

Zero bits are signaled by 180° phase reversals:



Example of a  
phase  
reversal

To avoid harmonics, the amplitude passes through zero at the instant of the reversal of phase. There is no need to change the amplitude when we don't change phase. Thus, we might see a signal amplitude like this:



# PSK31 signaling

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These phase changes occur every 32 milliseconds (31.25 Hz). Why this rate? Because:

- It can be easily derived from the 8KHz signal generated by your sound card (8000 Hz divided by 256 = 31.25 Hz), and
- 30 or so bits per second translates to about 50 wpm, the fastest you can type.

A zero (space) is defined as a 180 degree phase reversal from the prior sample,

A one (mark) is defined as no phase change from the prior sample.

# PSK31 signaling



Phase shift by 180, bit is a '0'

Phase shift by 180, bit is a '0'

Phase unchanged, bit a a '1'

Phase shift by 180, bit is a '0'

Phase shift by 180, bit is a '0'

Therefore, this signal is detected as 00100, which we will soon see is the representation of the character "SPACE".

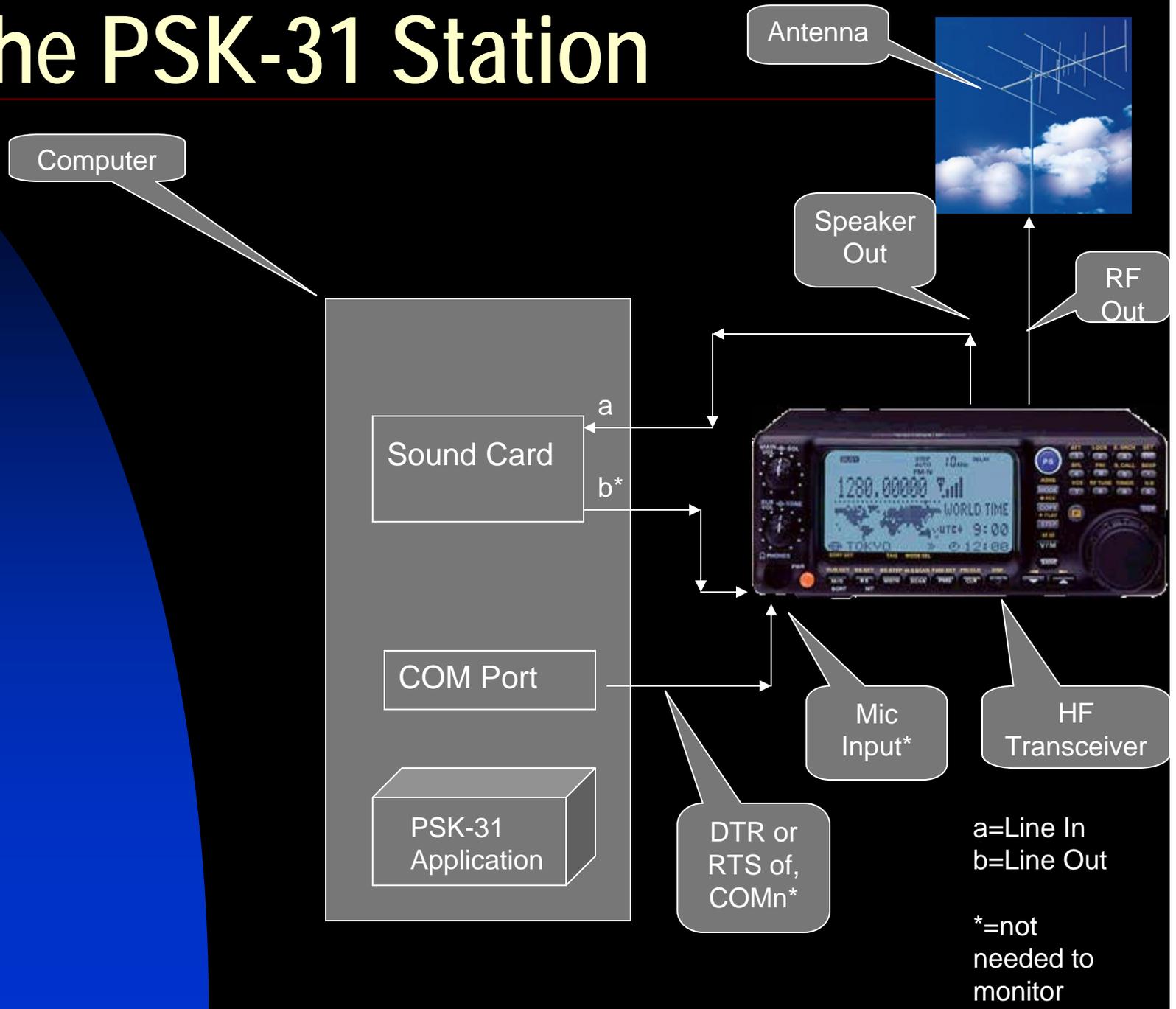
# PSK31 signaling

Let's look at PSK31 in the real world on an oscilloscope:



Notice that when we shift the signal one bit width (32 msec), and compare, the phase shift of 180 degrees is clear. This also gives us a clue for how PSK-31 is detected and converted back to the digital world.

# The PSK-31 Station



# Tuning up in PSK-31

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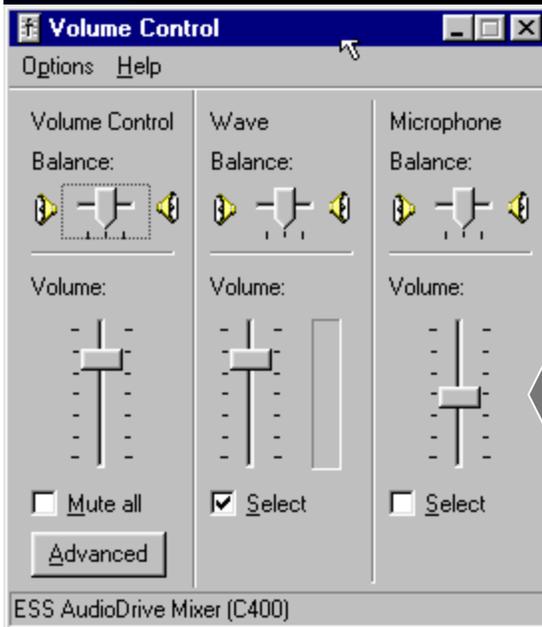
- Improperly tuned PSK-31 can result in
  - ◆ Undermodulation—your signal will be too weak to be received or copied
  - ◆ Overmodulation—your signal will have high Intermodulation Distortion (IMD) and splatter, and possibly will have poor copy also.

# Tuning up in PSK-31

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- Tuning up involves setting
  - ◆ Transceiver RF frequency
  - ◆ Transceiver microphone gain
  - ◆ Transceiver power setting
  - ◆ Turning off the speech processor
  - ◆ Turning off AGC
  - ◆ *Sound card line out audio level*

# Setting Audio Levels



- It's all in the Window's "mixer."
- At the mixer's main display...
- De-select everything but "Wave" (that's your line out).
- Set the Wave volume to maximum.
- Now you can adjust the sound level by the Volume Control slider.

# Setting the audio level

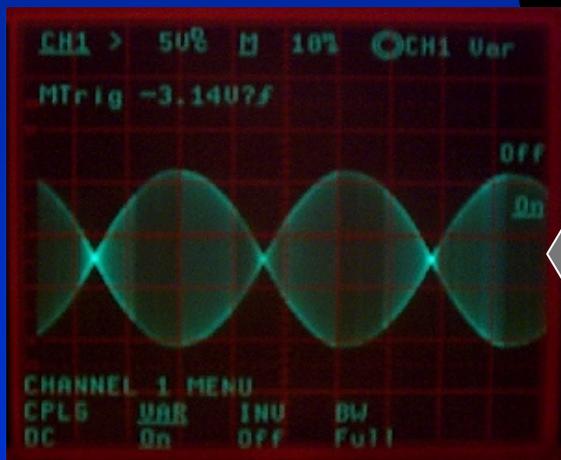
- Click “TX” button of your PSK-31 software, so that you are sending an idle PSK signal to the transceiver.
- Starting with the maximum sound level, observe the transmitter’s power output and back off the mixer’s audio level until the power level is at about 50%.
- You should observe an S-shaped curve like the one I measured at 1500 Hz.
- The proper audio level in this case would be about 50% of maximum.



# Setting the audio level



- Avoid the temptation to crank up the audio level to get more power!
- You will only produce horrendous splatter, have very poor IMD, and be less copiable!

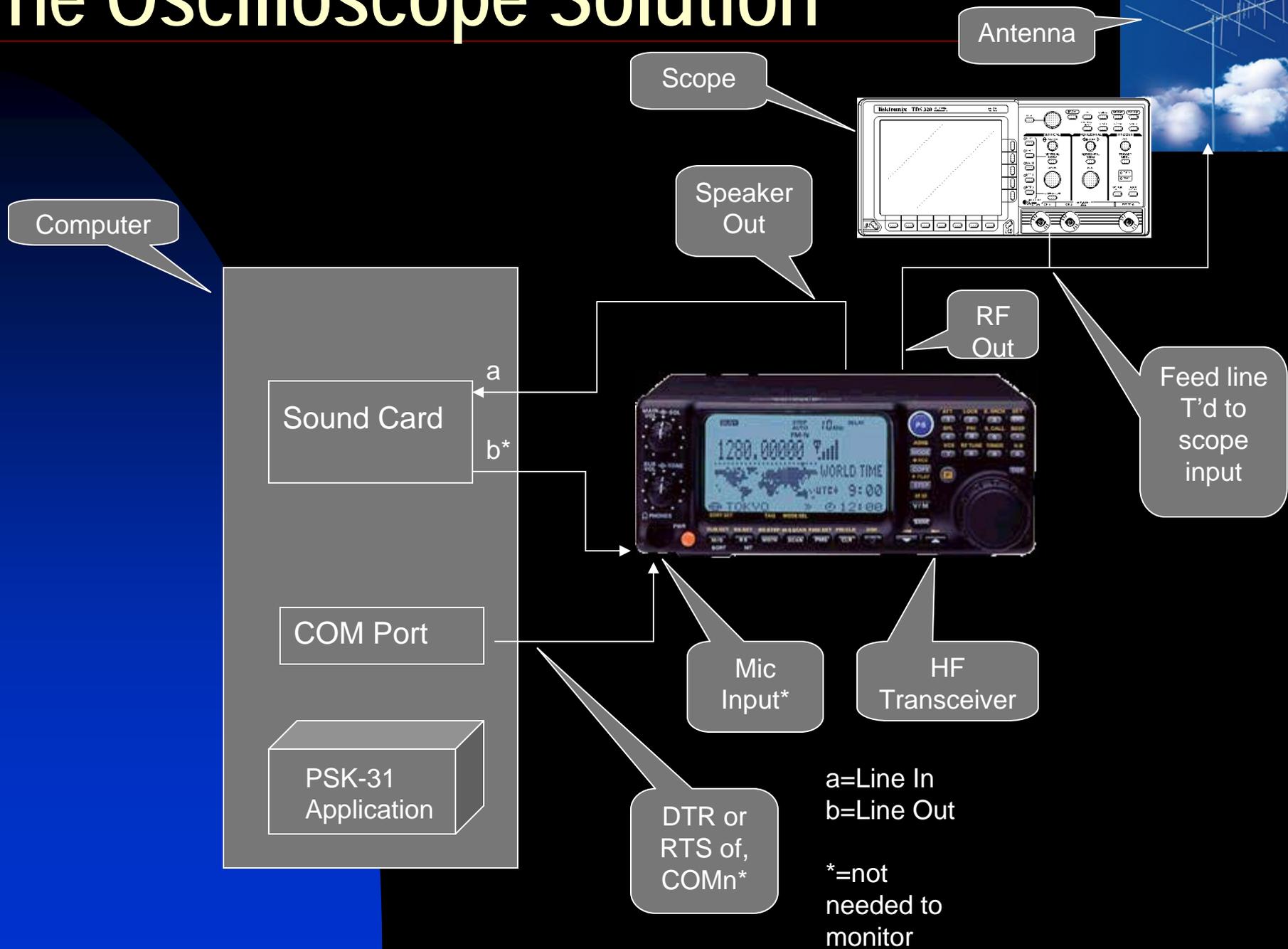


- Properly tuned, you will have a low IMD figure (-25 to -30 dB), and a clean spectrum with maximum copy—a signal you can be proud of.

# Setting the Audio Level is *not* a one-time Adjustment!

- The ideal audio level is a function of audio frequency. What works at 1500 Hz will not be a good setting at 300 Hz or 2500 Hz.
- Your rig's audio frequency response is not flat over the audio range but peaks in the mid-range from 1000-2000 Hz.
- Audio levels set by your Windows mixer will change as you use other applications.

# The Oscilloscope Solution



# What's the downside?

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- Poor use of a costly resource
- Bulky, power consumptive (not a good solution for a portable PSK-31 station)
- No place to get a good trigger signal
- Still a manual operation (you fiddle with the volume control slider).

# An alternate solution: let your computer be your scope

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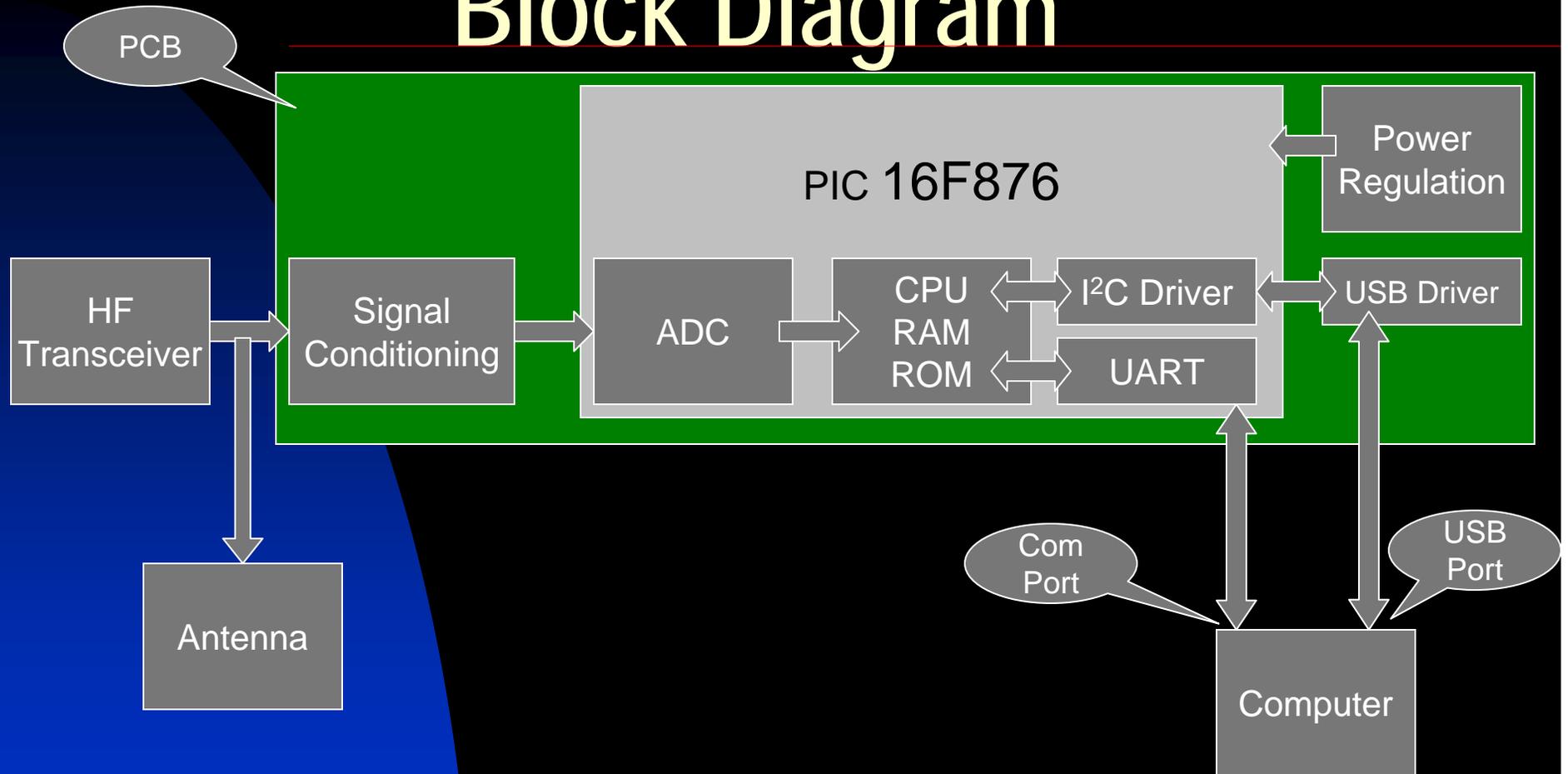
- A computer with display is guaranteed to be available
- Automatic graphical interface (if running a Windows OS)
- Only need to add an Analog to Digital Converter (ADC)
- Can do more than monitor RF— can dynamically set your audio for perfect output!

# Adding an ADC

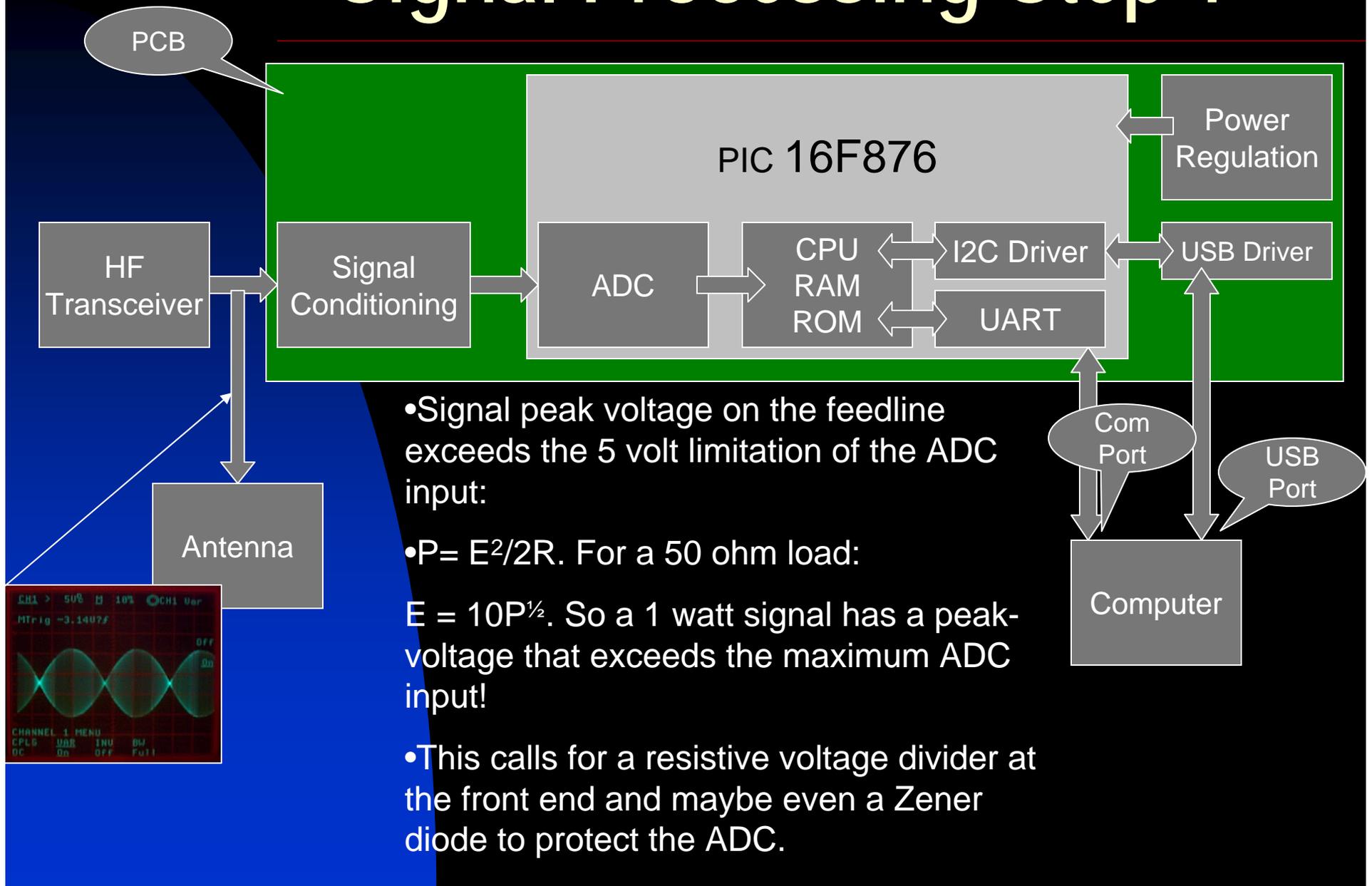
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- Let's use a PIC microcontroller as an inexpensive way to add an 8-bit ADC to your computer.
- The PIC 16F876 comes with a built-in UART that can communicate with your computer's COM port
- Has a built in I<sup>2</sup>C (2 wire) bus interface, so you can
  - ◆ Add more memory
  - ◆ Interface to a USB chip that can communicate with your computer's USB port

# Component Block Diagram

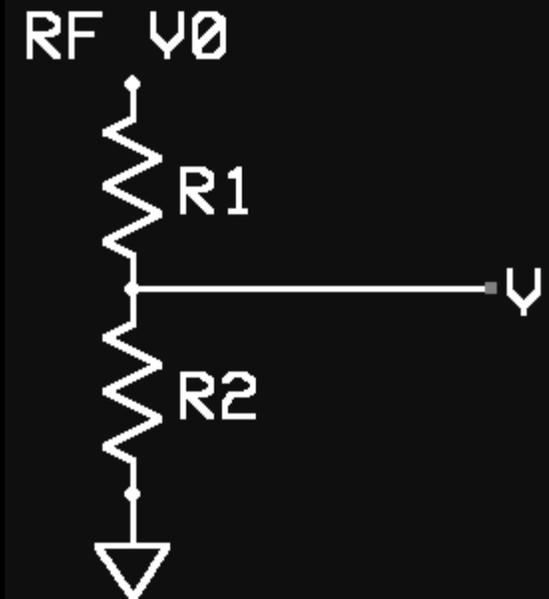


# Signal Processing-Step 1

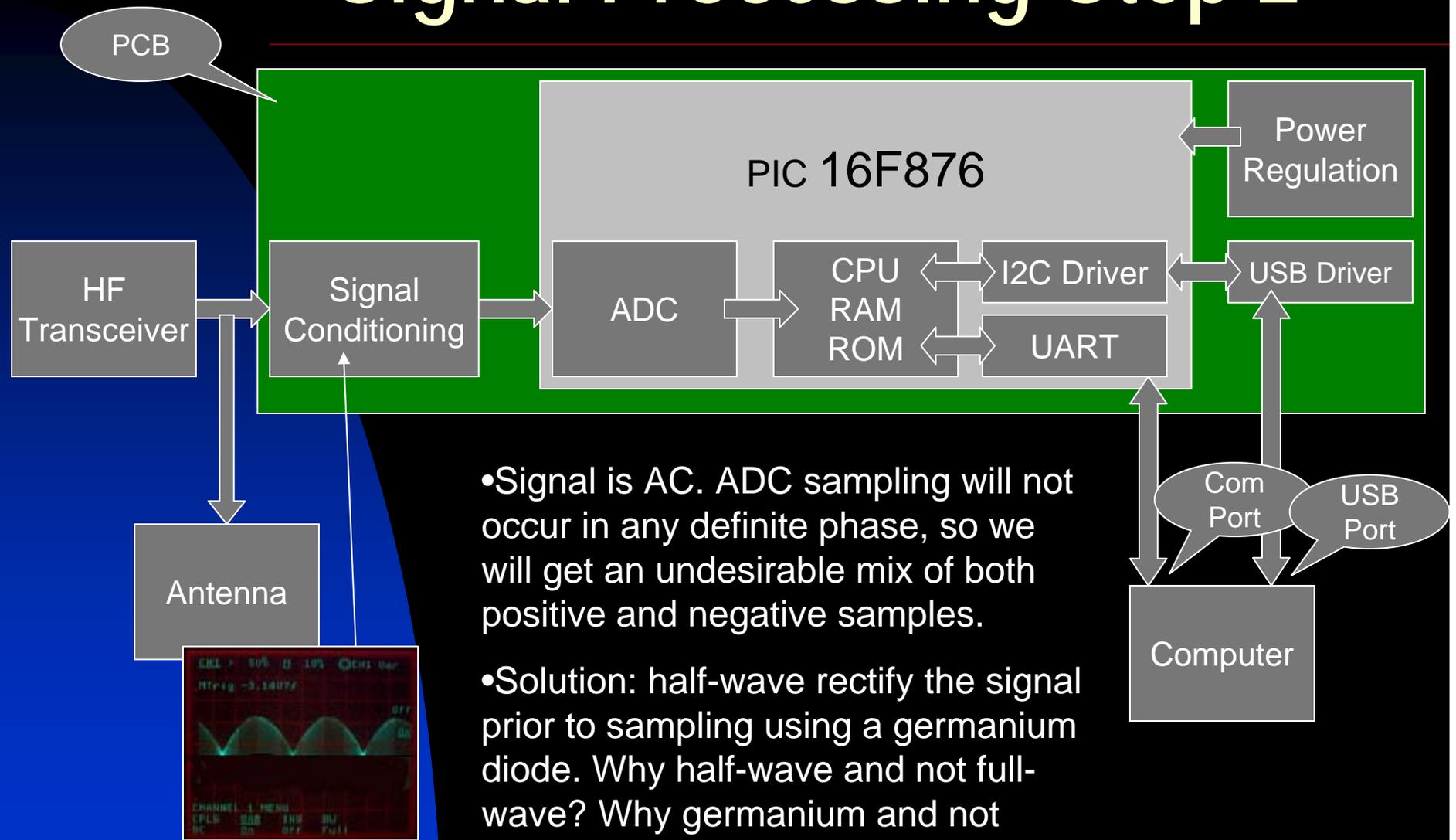


# Signal Processing-Step 1

- Peak RF voltage is given by
- $V_0 = (2R_A P)^{1/2} = 10P^{1/2}$  where  $R_A = 50\Omega$ , the antenna impedance
- Voltage applied to the ADC will be
- $V = V_0 R_2 / (R_1 + R_2)$
- Setting  $V=5$  (the ADC limit), substituting for  $V_0$  and solving for  $R_2 / R_1$  gives:
- $R_1 / R_2 = 2P^{1/2} - 1$
- Examples: if  $R_2 = 2\text{ K}\Omega$  then  
 $P = 4$  watts,  $R_1 = 6\text{ K}\Omega$   
 $P = 100$  watts,  $R_1 = 38\text{ K}\Omega$



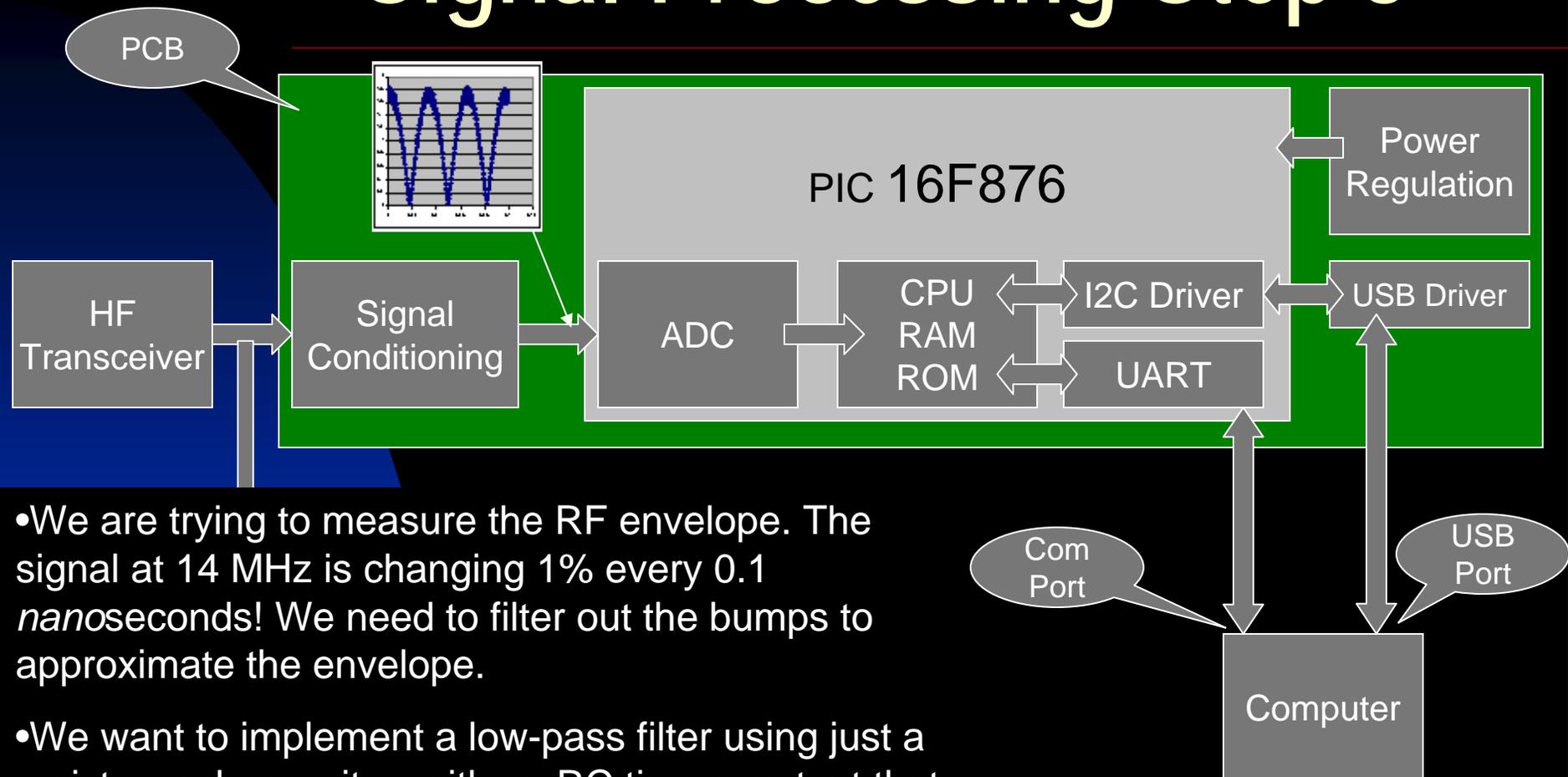
# Signal Processing-Step 2



- Signal is AC. ADC sampling will not occur in any definite phase, so we will get an undesirable mix of both positive and negative samples.

- Solution: half-wave rectify the signal prior to sampling using a germanium diode. Why half-wave and not full-wave? Why germanium and not silicon?

# Signal Processing-Step 3



- We are trying to measure the RF envelope. The signal at 14 MHz is changing 1% every 0.1 *nanoseconds*! We need to filter out the bumps to approximate the envelope.

- We want to implement a low-pass filter using just a resistor and capacitor, with an RC time constant that allows the 31 Hz signal to pass, but blocks the RF.

- The PIC's ADC has an internal input capacitance of 5 pf. So the question is: how to determine the value of the resistor?

# Signal Processing-Step 3

The cutoff frequency in this low pass filter is

$$f_c = 1/(2 \pi RC)$$

And the “gain” of the filter is given by

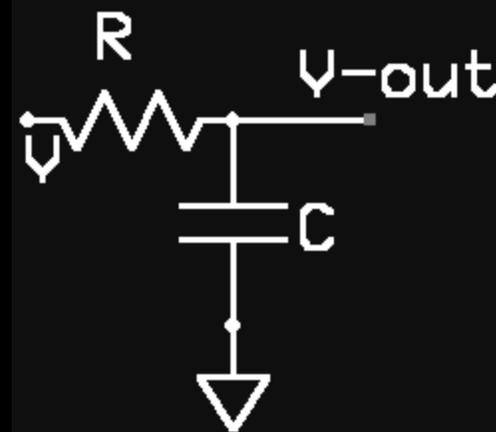
$$G = 1/(1 + (f/f_c)^2)^{1/2}$$

A good cutoff to choose would be the geometric mean between 31 Hz (the signaling envelope) and 14 MHz (the carrier frequency). This works out to be 21 KHz.

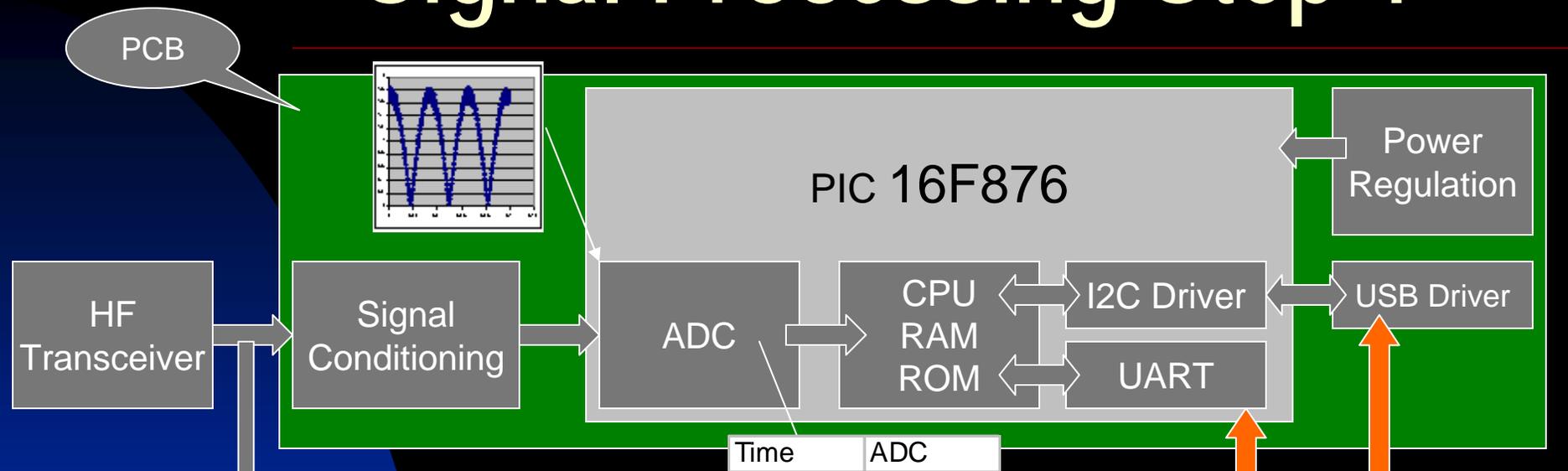
Then G for the RF component is  $1.5 \times 10^{-3}$  and for the PSK31 component is  $1 - 10^{-6}$ .

Taking a value of C to be 0.1 uf, we have

$$R = 1/(2 \pi f_c C) = 113 \text{ Kohm} \sim 100\text{K}$$



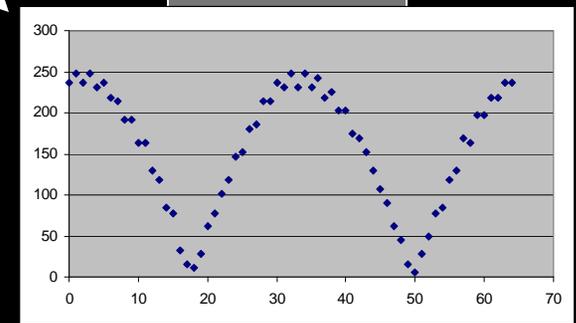
# Signal Processing-Step 4



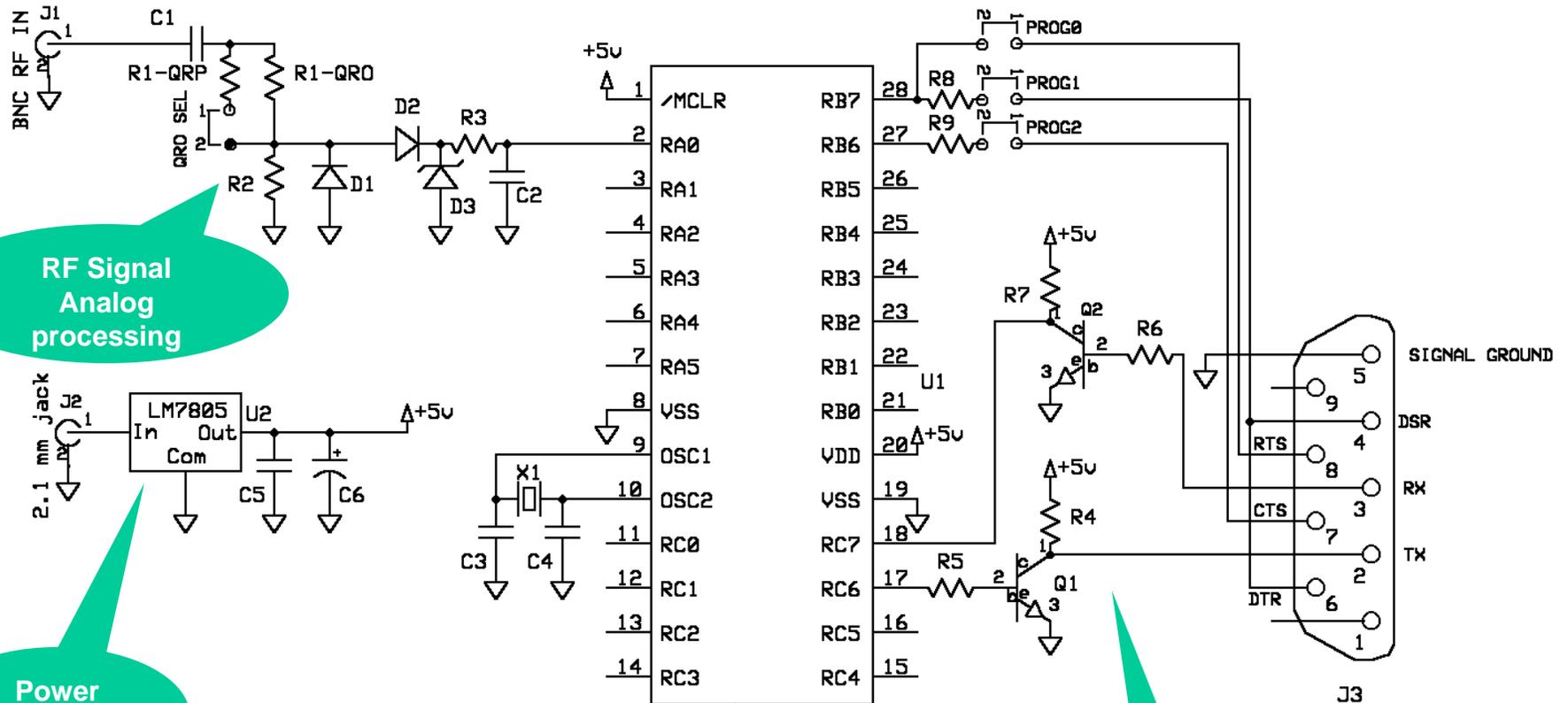
•The firmware now waits for a request from the computer, then instructs the ADC to sample the signal, stores the digital result in a table in RAM memory, then transfers the buffered bytes to the computer.

•How often to sample? Recall that the PSK31 envelope has a frequency of 31.25 Hz (1 bit or cycle per 32 msec). By sampling the waveform every millisecond we obtain exactly 32 samples for each PSK31 bit. In 64 msec, we can acquire 2 PSK31 bits in 64 samples.

Time	ADC
0	236
1	248
2	236
3	248
4	231
5	236
6	219
7	214
8	191
9	191
10	163
11	163
12	129
13	118
14	84
15	78



# PSK Meter Circuit Diagram



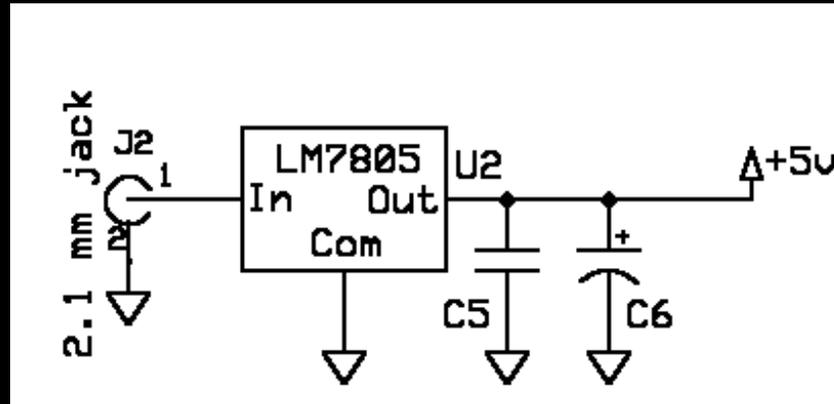
RF Signal  
Analog  
processing

Power  
supply

PIC  
Microcontroller

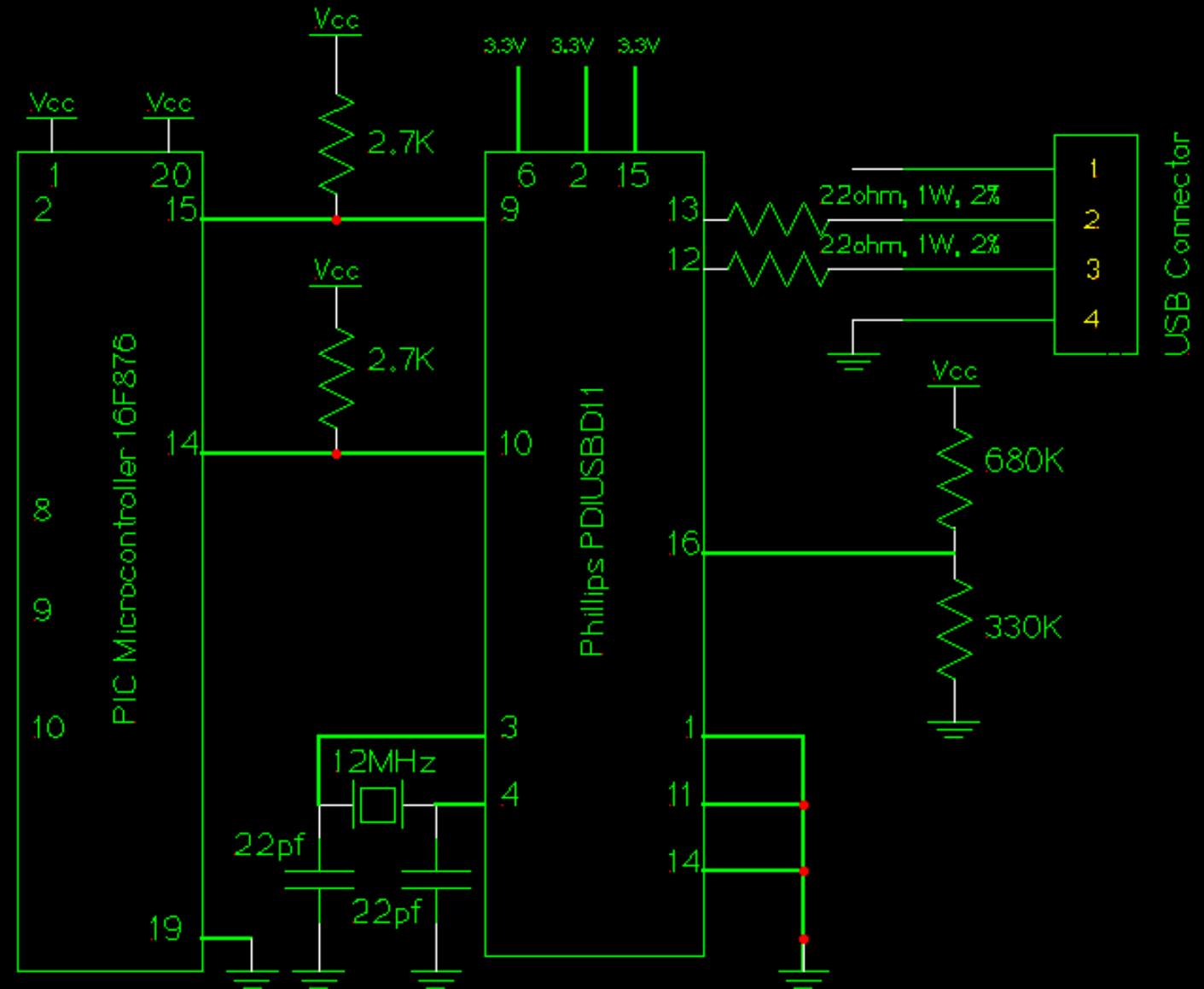
Serial I/O to  
computer

# Adding the Power Supply Circuit

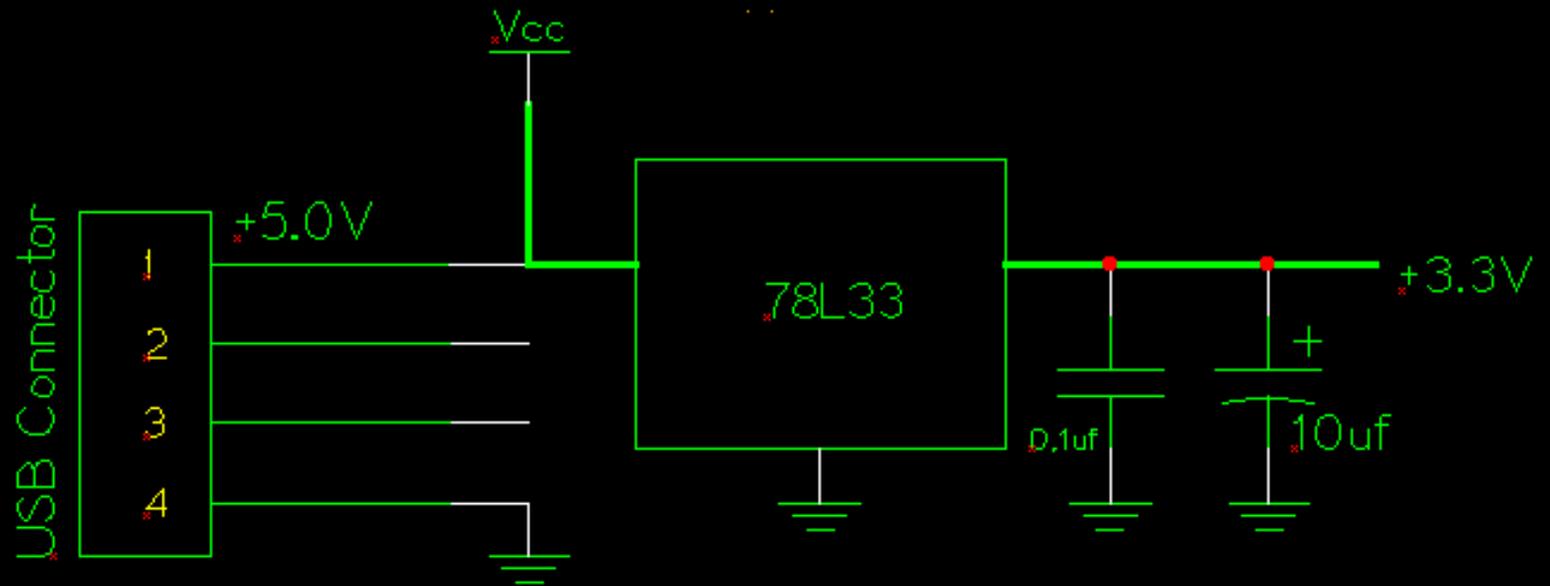


- Accepts any positive voltage from 7.5 to 35 volts and outputs +5.0 volts. Package: TO-92.

# PSK meter with USB interface



# Revised Power Circuit for USB



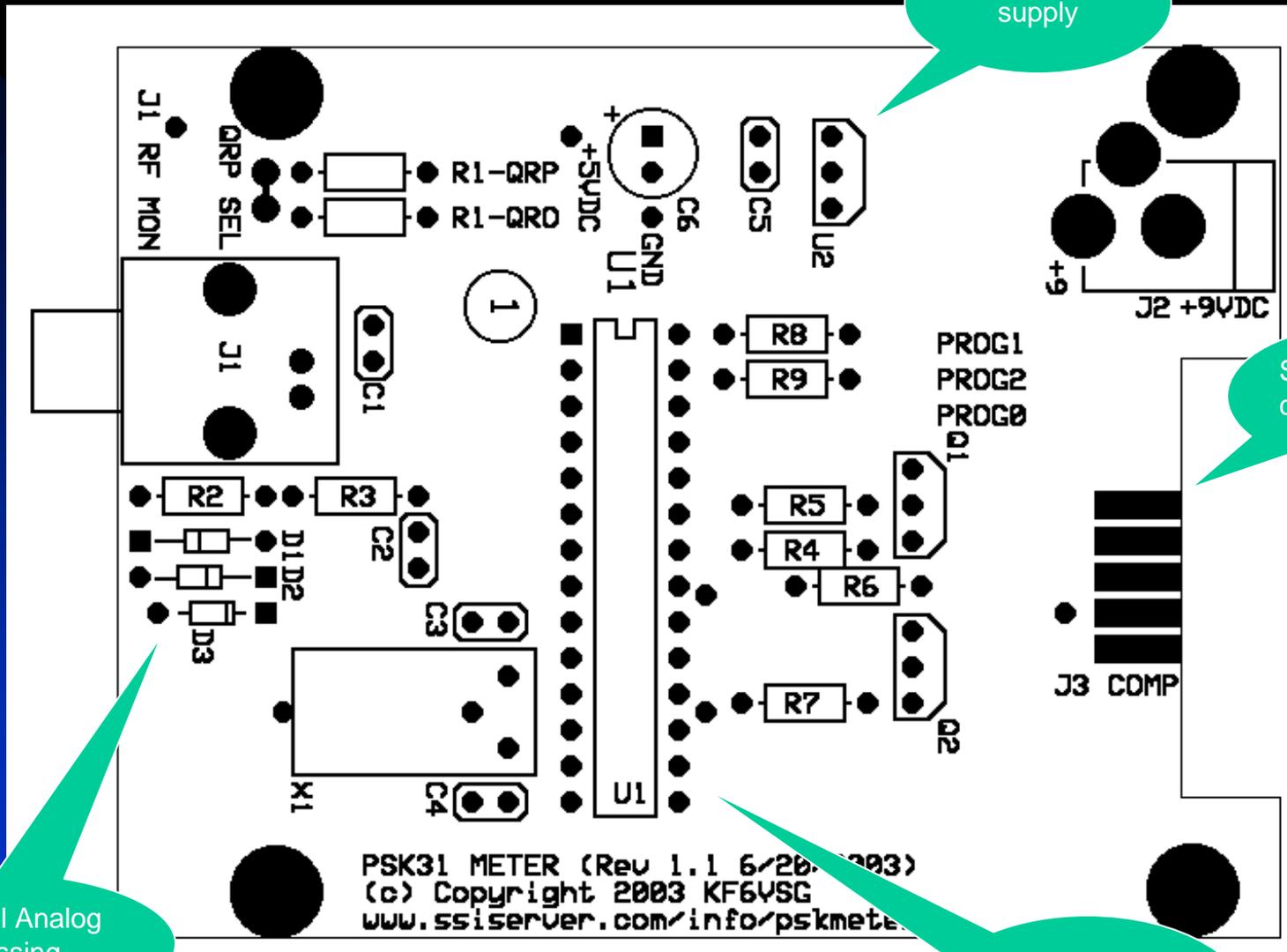
- Other enhancements:
- Option for either RS232 or USB interface on the same card
- Diode isolation between USB power and external power
- On board PIC programming

# PCB Layout

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- In theory the circuit should and does work, *but...*
- You have RF signals and digital signals on the same board...so *what?*
- Strong RF signals at 14 MHz will mix with the 20MHz crystal and shift the CPU's clock, and the firmware will crash
- RF should be isolated from any signals that could creep into your computer, or you will have software problems
- The solution is to *be careful with the physical layout* of the board:
  - Keep all RF traces as short as possible.
  - Use plenty of ground plane on the copper side of the board, and even the component side also.
  - Keep the PIC and its crystal away from the RF signal conditioning components and traces.
  - Use ferrite beads (toroids) on the cables to your computer.

# PCB Layout



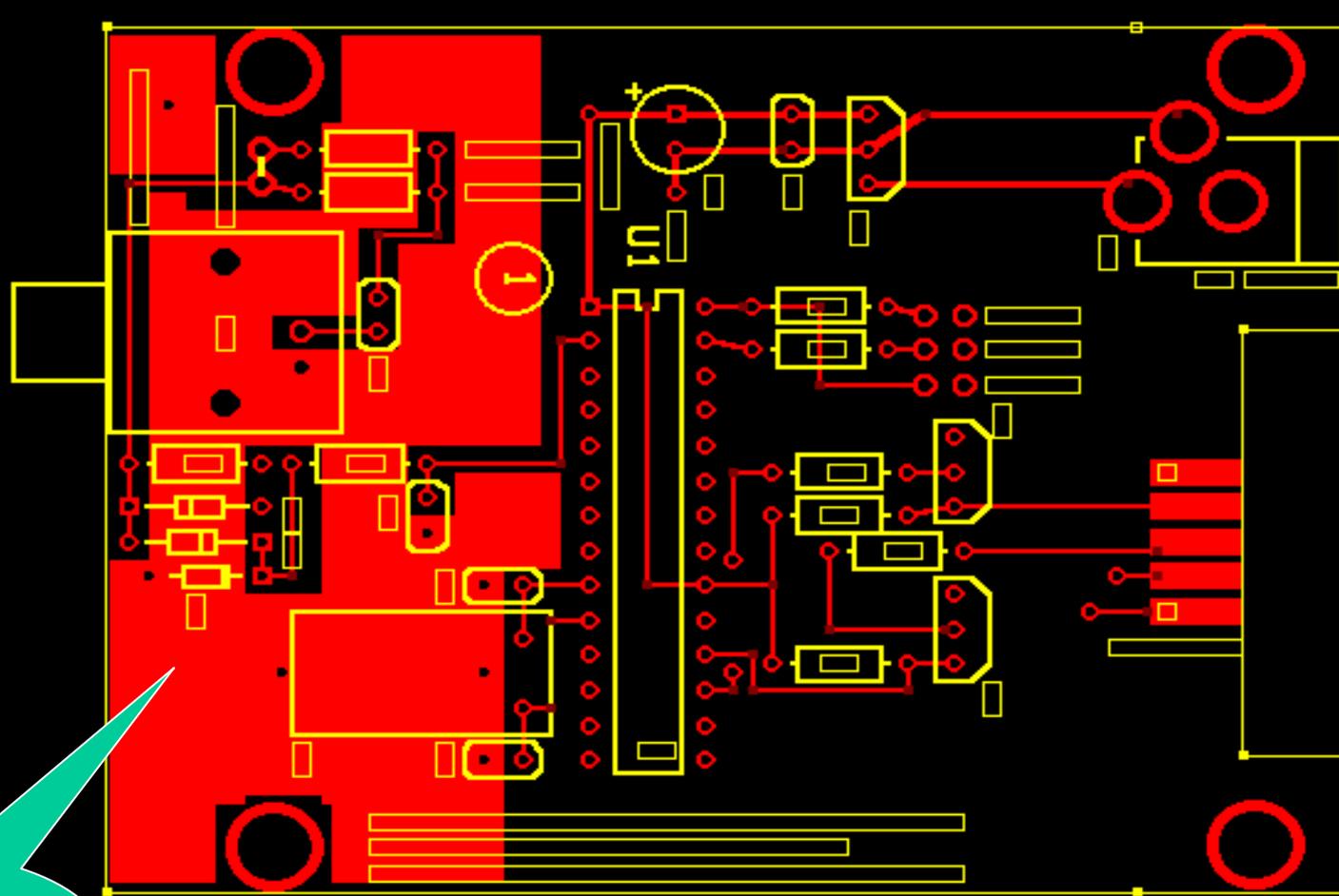
Power supply

Serial I/O to computer

RF Signal Analog processing

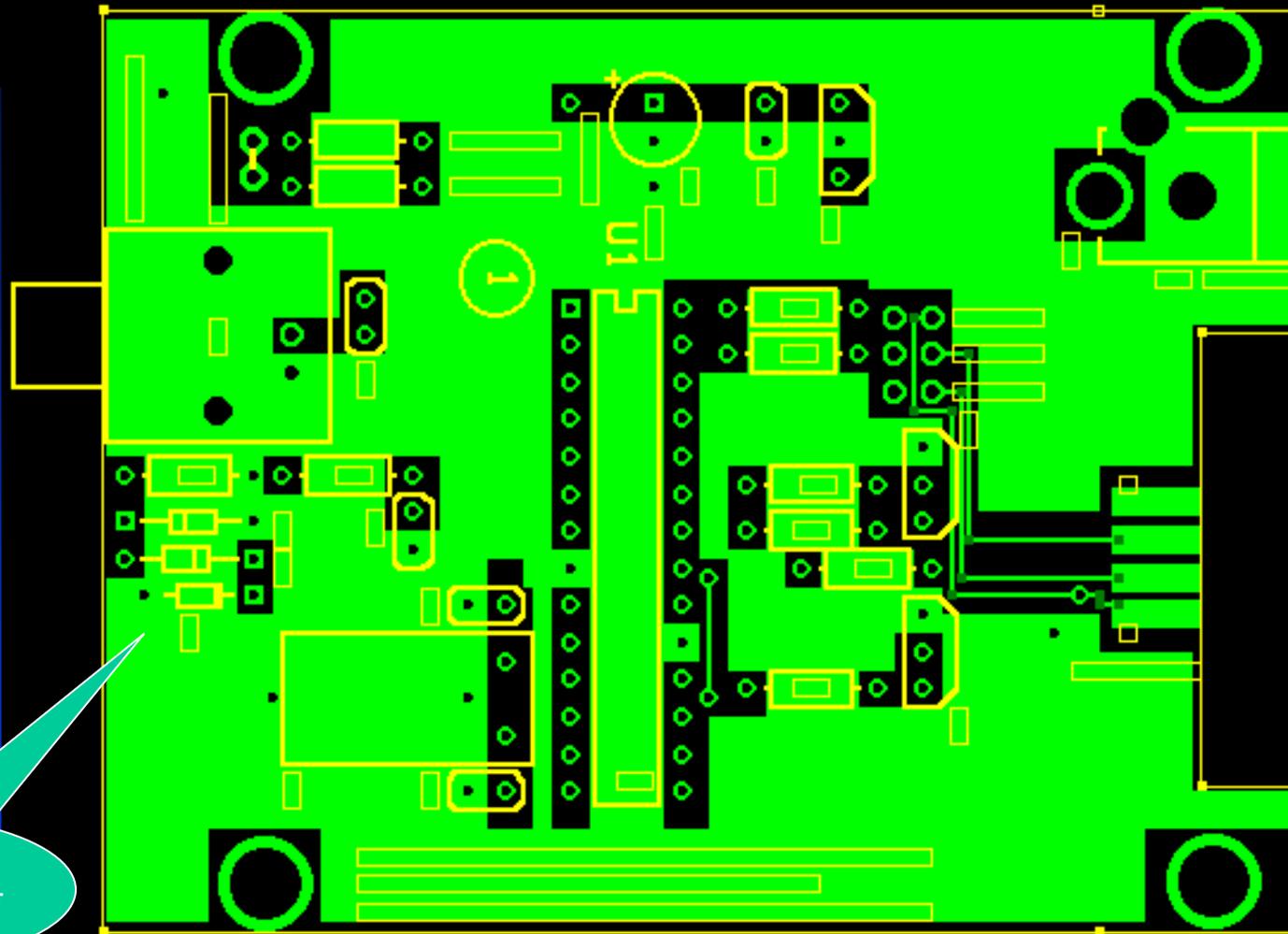
PIC Microcontroller

# Component Side Ground Plane



Red indicates  
component side  
copper ground  
plane

# Solder Side Ground Plane



Green indicates  
solder side copper  
ground plane

# PSK Meter Printed Circuit Board

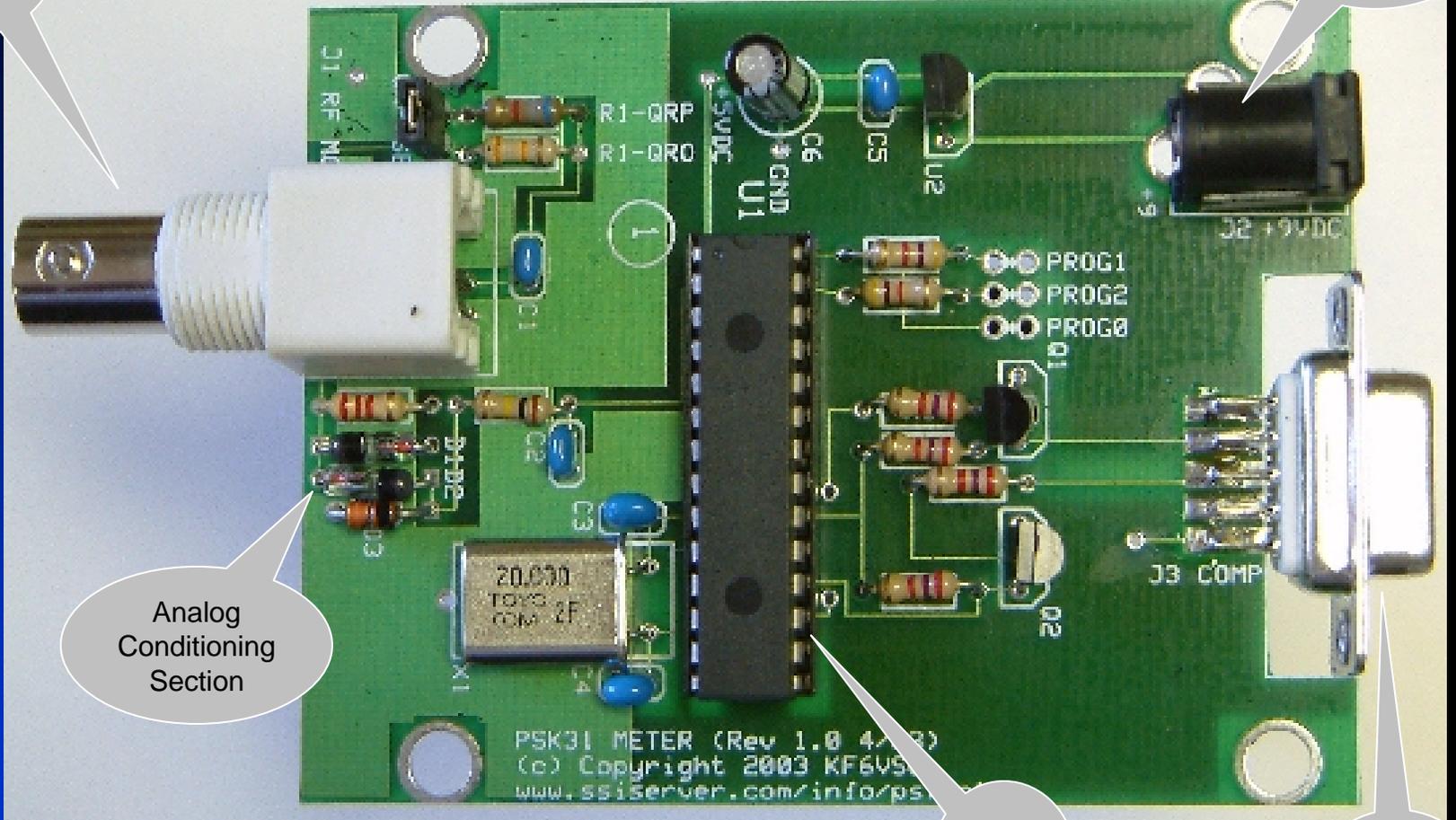
RF  
Input  
BNC

Power  
Supply  
Section

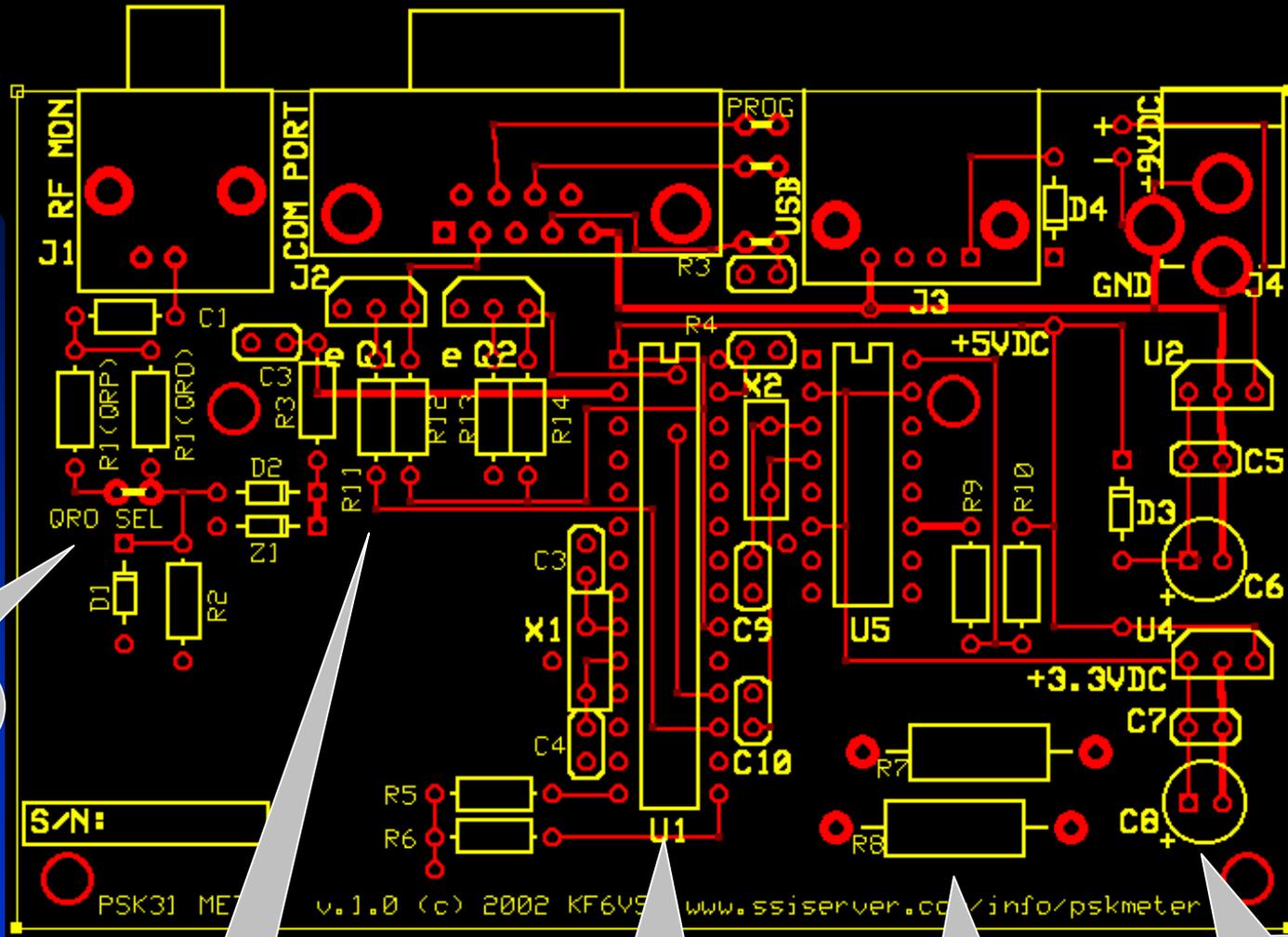
Analog  
Conditioning  
Section

PIC  
MPU

RS232  
to  
COM  
port



# PSK Meter Printed Circuit Board



Analog Conditioning Section

RS-232 Section

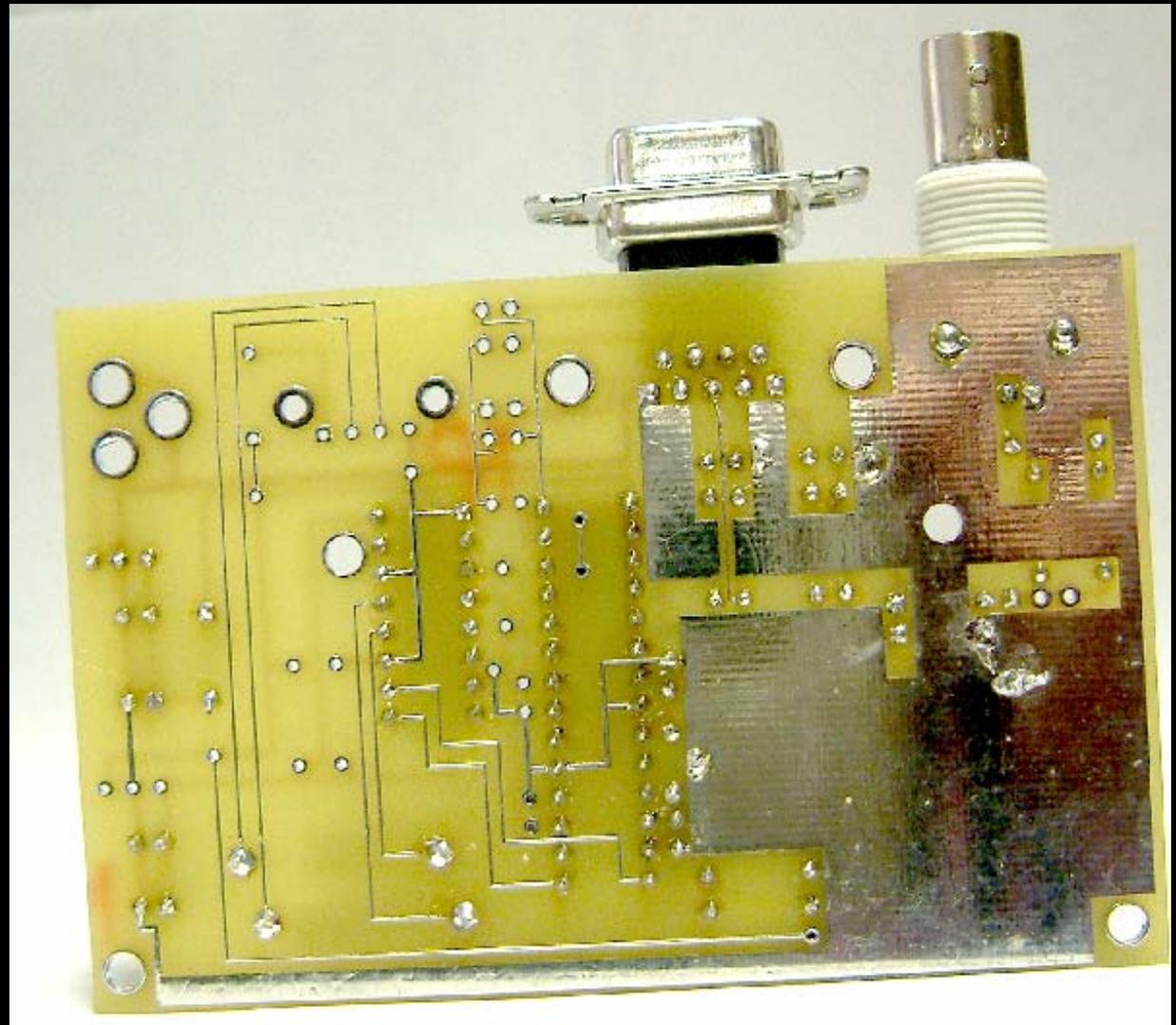
Microprocessor Section

USB Section

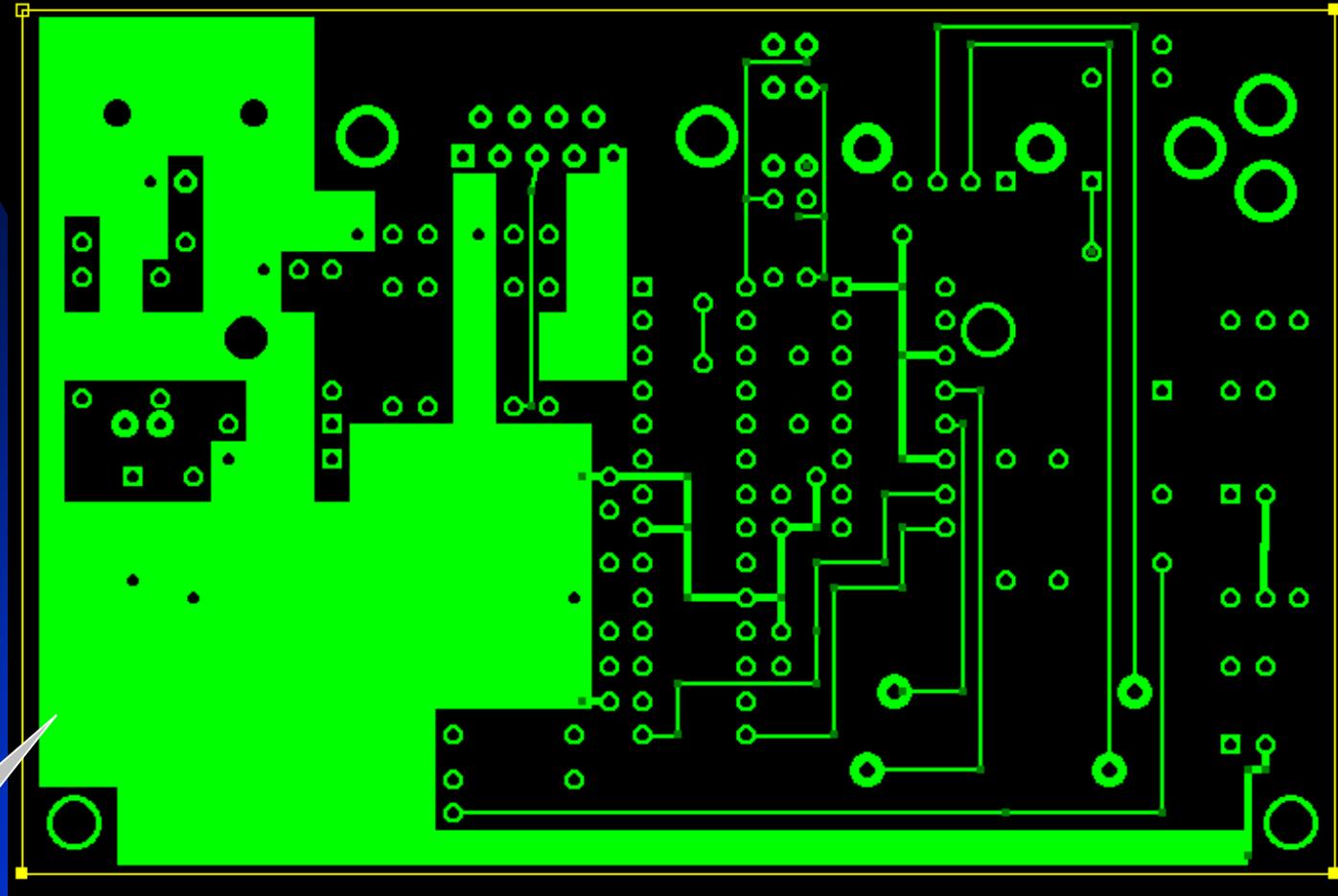
Power Supply Section

# PSK Meter Printed Circuit Board

- Solder, partially assembled
- Prototype board is missing silk screen and solder mask
- Notice lots of ground plane used at RF input and signal conditioning section. This is a must for the RF to stay out of the MPU!



# PSK Meter Printed Circuit Board



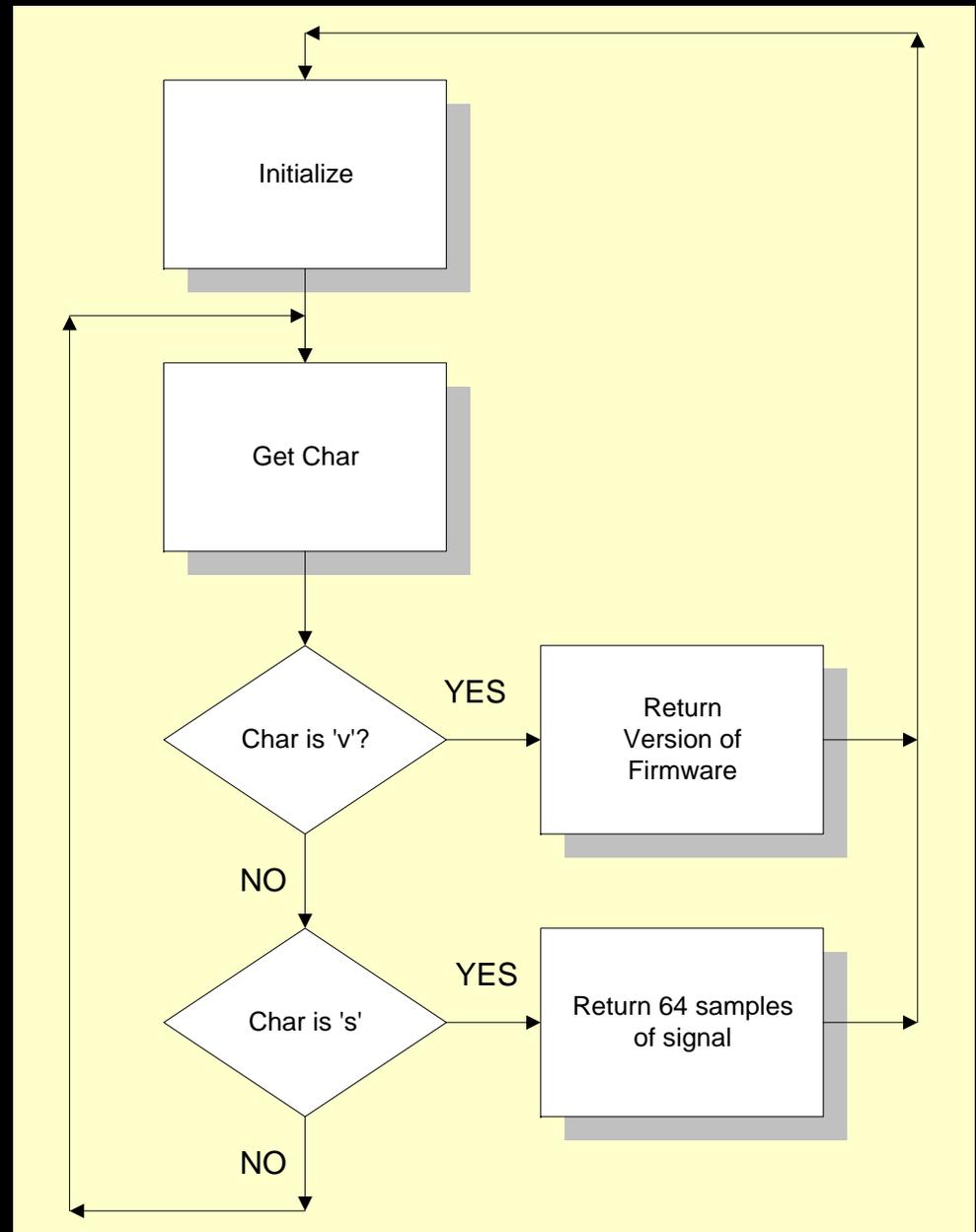
Ground  
Plane

# Parts List—COM Port version

Item	Count	Circuit Part Number	Description
1	1	R1-QRP	7K $\Omega$ ¼ watt 10%
2	1	R1-QRO	38K $\Omega$ ¼ watt 10%
3	1	R2	2.2K $\Omega$ ¼ watt 10%
4	1	R3	1.5M $\Omega$ ¼ watt 10%
5	2	R4,R5	2.5K $\Omega$ ¼ watt 10%
6	2	R6,R7	1.0K $\Omega$ ¼ watt 10%
7	2	C1,C2	0.1uf
8	2	C3,C4	22 pf
9	1	C5	10uf electrolytic
10	2	D1,D2	Germanium diodes 1N34A or OA47 or NTE109 or ECG109 or SK3090
11	1	Z1	4.7V ½ watt Zener diode, 1N4624 or NTE5009A
12	2	Q1,Q2	2N4401 NPN transistor
13	1	X1	20 MHz xtal, Mouser part 73-XT49U2000-20
14	1	U1	PIC 16F876
15	1	U2	7805 5V reg, TO-92
16	1	28 pin DIP socket	
17	1	J1	BNC edge connector, Amphenol 31-5431-2010
18	1	J2	DB9 female edge connector, Amphenol 745781-4, Mouser part 152-3409
19	1	J4	Power jack, 2.1 mm, edge connector, Digikey CP-202A-ND,JDR # PC21S

# Firmware

- Firmware: software that runs on the PIC in ROM
- Initializes hardware and variables
- Waits for a character command from the PC
- Returns either
  - ◆ information about itself, or
  - ◆ takes 64 samples of the RF envelope at 1 msec intervals



# Firmware Source Code

```
#include <16f876.h>
#include <string.h>
#fuses HS,NOWDT,NOPROTECT
#use delay(clock=20000000)
#use rs232(baud=19200, xmit=PIN_C6, rcv=PIN_C7)

void showversion()
{
    printf("PSKMETER Firmware Version 1.00.\n\r");
    printf("(c) 2002 KF6VSG
           www.ssiserver.com/info/pskmeter\n\r");
}

void senddata()
{
    int i;
#define MAXSAMPLES 64 //exactly two psk31 cycles
    int sample[MAXSAMPLES];
// take MAXSAMPLES samples separated by 1 msec:
    for (i=0;i<MAXSAMPLES; i++)
    {
        sample[i] = Read_ADC();
        delay(960); // wait .96 millisecond
    }
// upload the samples to the windows app
    for (i=0; i< MAXSAMPLES; i++)
        printf("%c", sample[i]);
}
```

Copyright &  
Version Info

Take &  
report data

```
main()
{
    char command;
    showversion();
    setup_port_a( ALL_ANALOG );
    setup_adc( ADC_CLOCK_INTERNAL );
    set_adc_channel( 0 );

    do
    {
        /* wait for the command byte 's' or
           'v' to be received from the windows
           application */

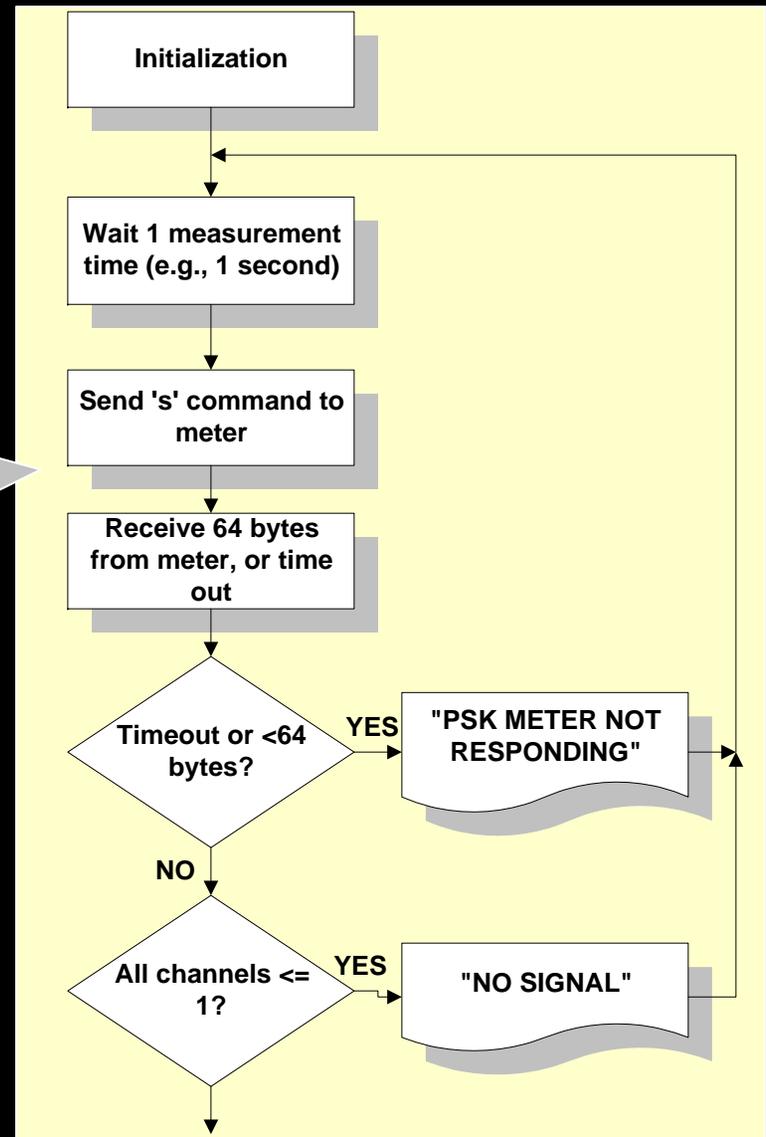
        do
            command = getc();
        while ((command != 's') &&
              (command != 'v'));

        if (command == 'v')
            showversion();
        else
            senddata();
    } while (TRUE);
}
```

Initialization

# PC Software Flow Diagram—Part 1

Loop waiting for a valid 64 byte envelope measurement



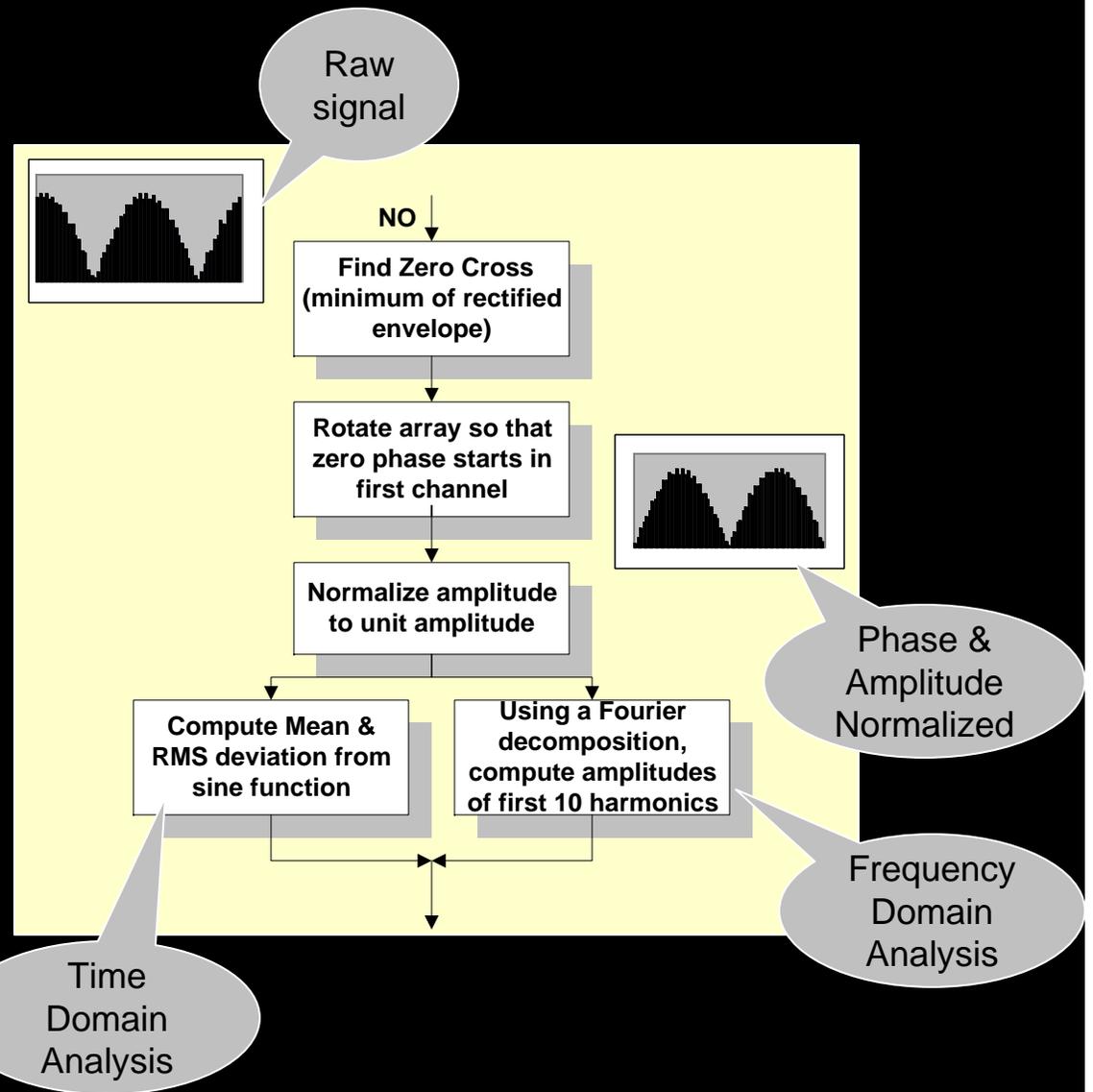
# PC Software Flow Diagram—Part 2

Computing RMS and Mean deviation in the *time domain*:

```
for i:=1 to 32 do begin
  error := s[i] - sin(theta[i]);
  rms := rms + sqr(error);
  mean := mean + error;
end;
rms := sqrt(rms)/(32);
mean := mean/32;
```

Computing harmonics in the *frequency domain*:

```
for j:=1 to 10 do
  for i:=1 to 64 do begin
    a[j] := a[j] + s[i]*sin(j*theta[i]);
    b[j] := b[j] + s[i]*cos(j*theta[i]);
  end;
```



# Metrics of Distortion

- RMS is a time domain metric of the departure of the signal from a perfect sinusoidal waveform. An RMS of zero indicates a perfect signal. Greater RMS values indicate clipping and other forms of distortion that result in splatter.
- Mean deviation is another time domain metric that is used to establish if the signal is overdriven (over-modulated) or underdriven (under-modulated). If the mean  $>0$ , the signal is over modulated. If the mean  $<0$ , the signal is under modulated.
- Software can make decisions regarding the quality of the signal based on these two metrics.
- However, we usually think in frequency space...

# Metrics of Distortion-Part 2

- The discrete Fourier coefficients,  $a_n$  and  $b_n$  are the real and imaginary amplitudes of the signal's fundamental frequency (31.25Hz) and its harmonics (62.5, 93.75, etc.)
- The absolute amplitude at each frequency is  $\sqrt{a_n^2 + b_n^2}$ .
- A perfect signal would consist only of the fundamental ( $n=1$ ). The amplitude of all the harmonics ( $n>1$ ) would be exactly zero.
- The amount of energy in the harmonics gives us the *Inter-modulation Distortion*, or IMD. Since the energy is proportional to the square of the amplitude, we have
  - $IMD = 10 \text{ Log}_{10}(H/E)$ .

```
for j:=1 to 10 do
  E := E + sqr(a[j])+ sqr(b[j]);
H := E - (sqr(a[1]) + sqr(b[1]));
F := H/E;
IMD := 10*log10(F);
```

# Metrics of Distortion—Part 3

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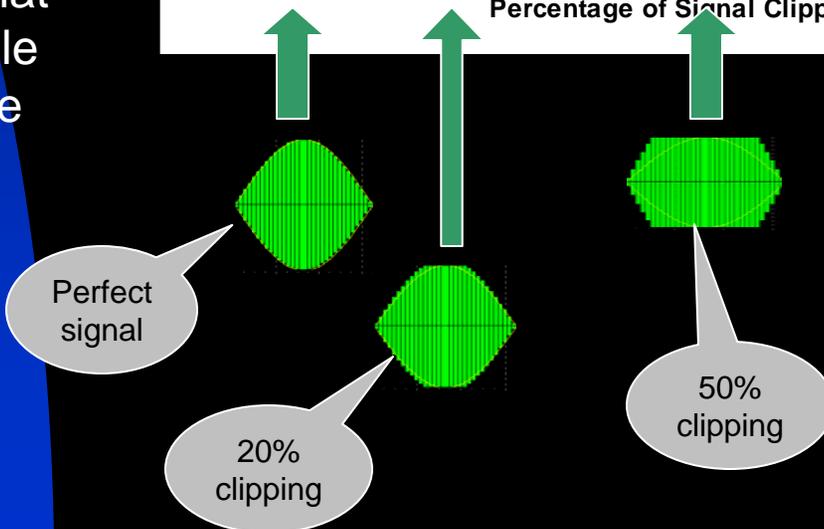
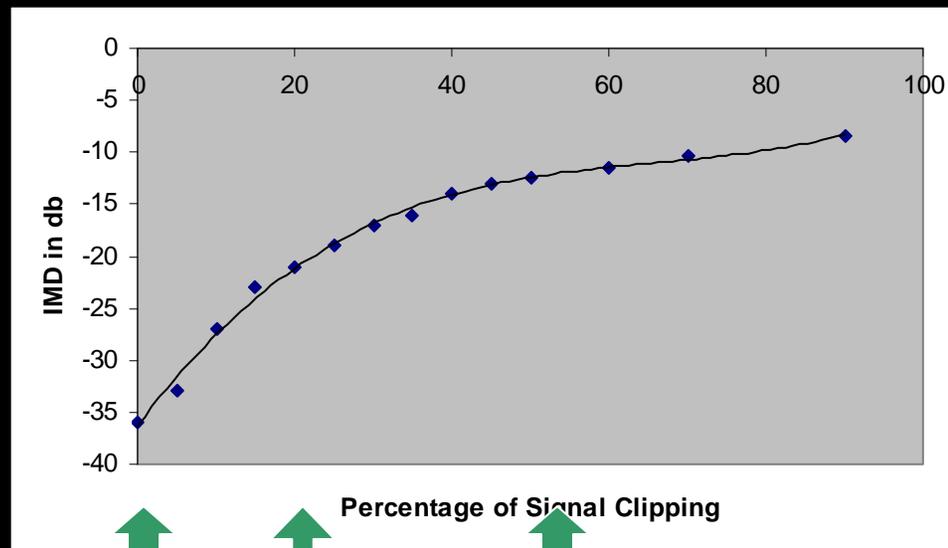
- $IMD = 10 \text{ Log}_{10}(H/E)$
- Since the energy in the harmonics will always be less than the total energy,  $H/E$  will always be a fraction, so IMD will always be negative.
- IMD is quoted in units of dB.
- Large absolute IMD is better. Smaller absolute IMD is worse.

# Example: IMD vs. Clipping

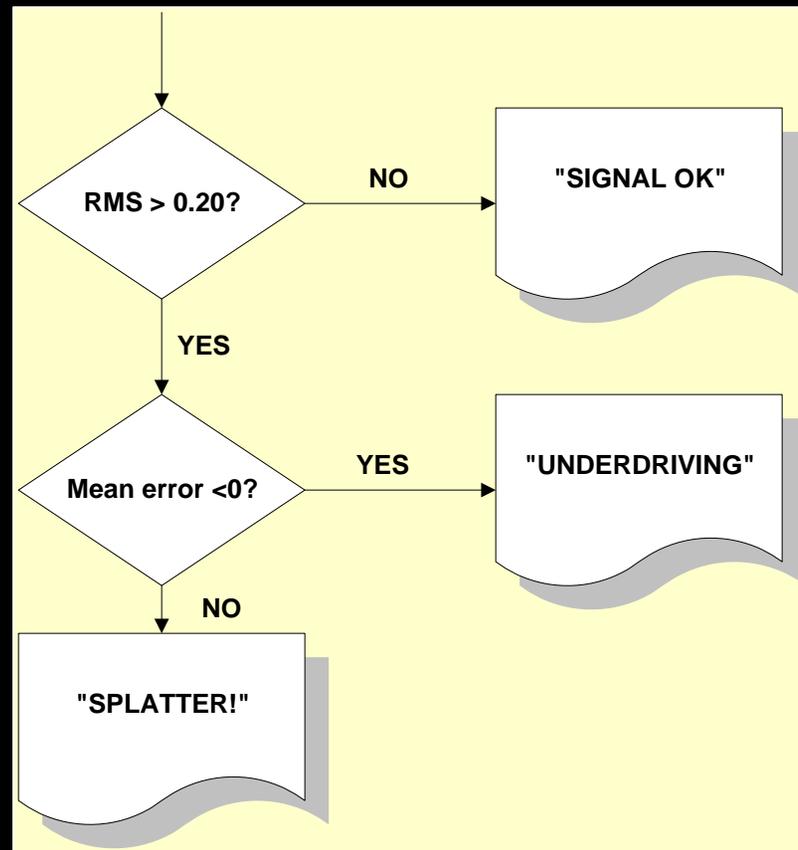
To create this example, I simulated a PSK31 signal with a small amount of white noise, and then computed the IMD from the Fourier analysis for various degrees of “clipping”, i.e., flat-topping the signal.

In the real world, an IMD is  $-30\text{db}$  is excellent. Don't expect to do better. An acceptable IMD (one that won't result in noticeable splatter) would be in the range of  $-20$  to  $-30$ .

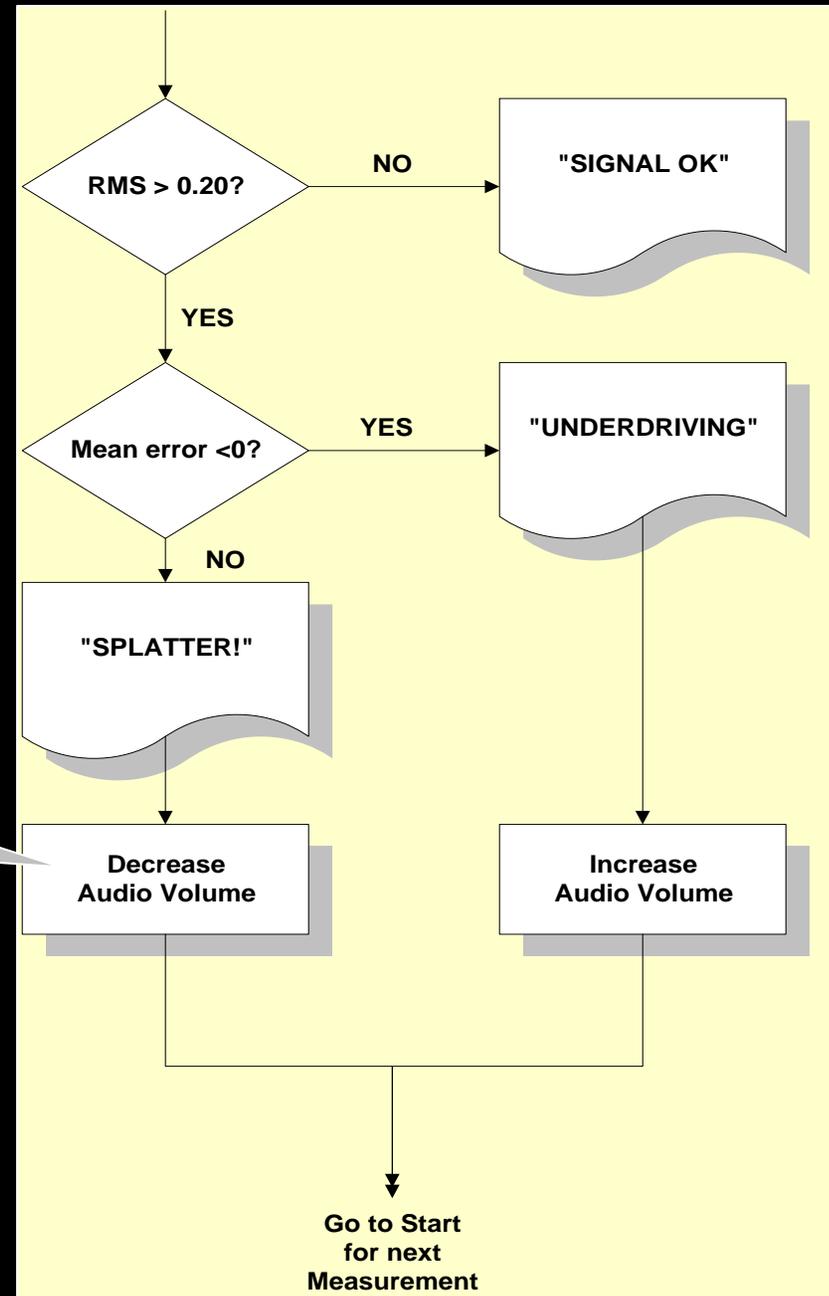
If you are getting IMD reports of  $-10$ , you are splattering all over the band!



# PC Software Flow Diagram—Part 3



# PC Software Flow Diagram—Part 4



Automate the system:  
Add a proportional negative feedback loop!

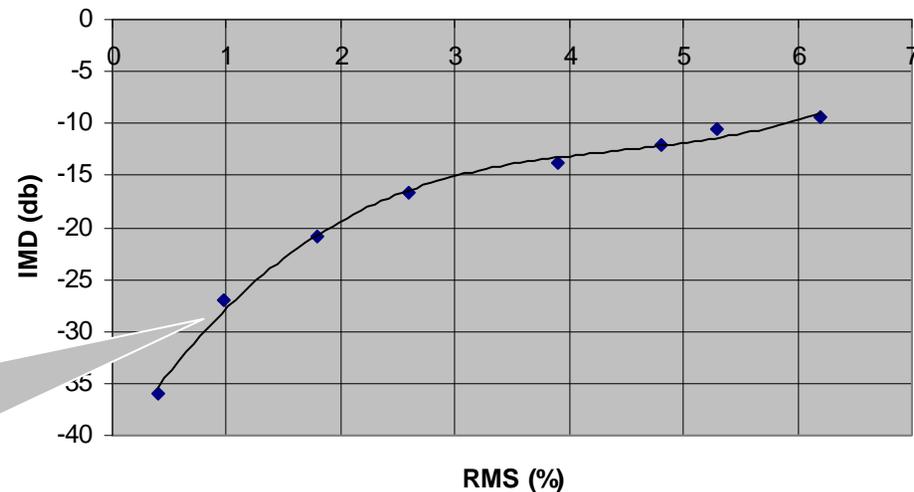
# Monitor and Control

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- The PSK Meter monitors your RF output and provides a visual, scope-like display to confirm that your signal is clean.
- However, the software also controls the audio level of your sound card to obtain the largest signal without splatter—automatically!

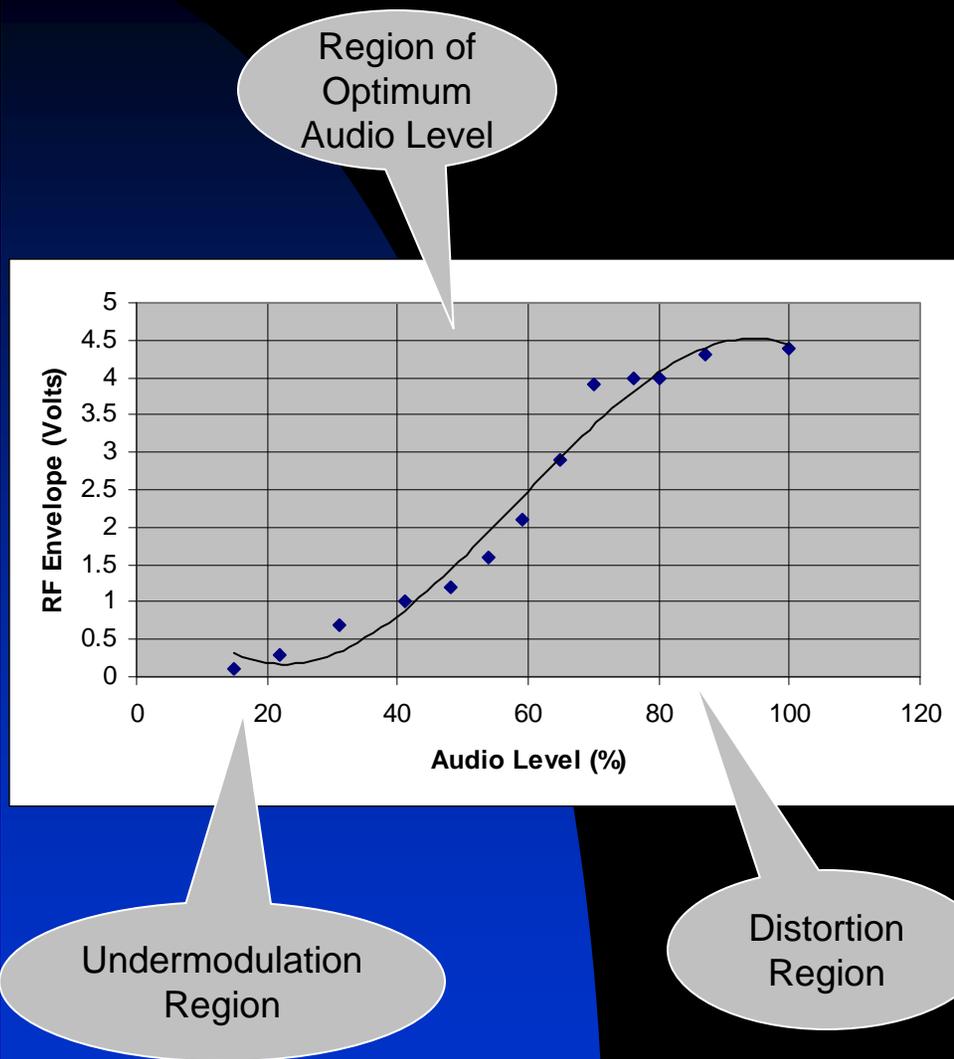
# Control variables—what to use?

- RMS (deviation from ideal signal), in percent
- IMD (harmonic distortion), in dB
- Average deviation—discerns between overmodulation and undermodulation



In most cases, RMS and IMD can be used interchangeably as the control variable

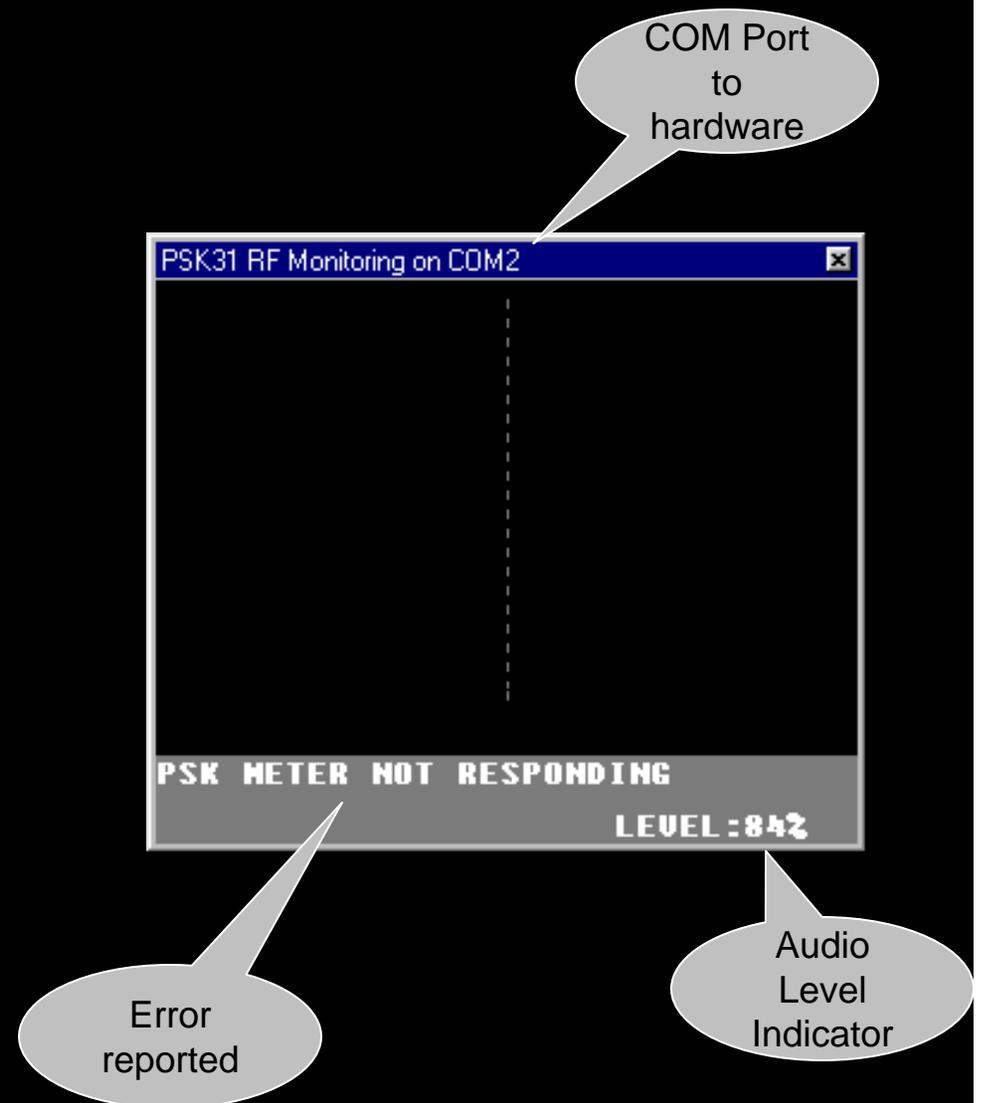
# A graphical look at feedback



- Plot the peak PSK31 envelope vs audio level.
- Notice that clipping occurs for the top 1/3 in audio level.
- Define underdriving as the bottom 1/3 in audio level.
- The feedback loop in the software attempts to find and place the audio level in the middle third, tending to the largest signal without distortion.

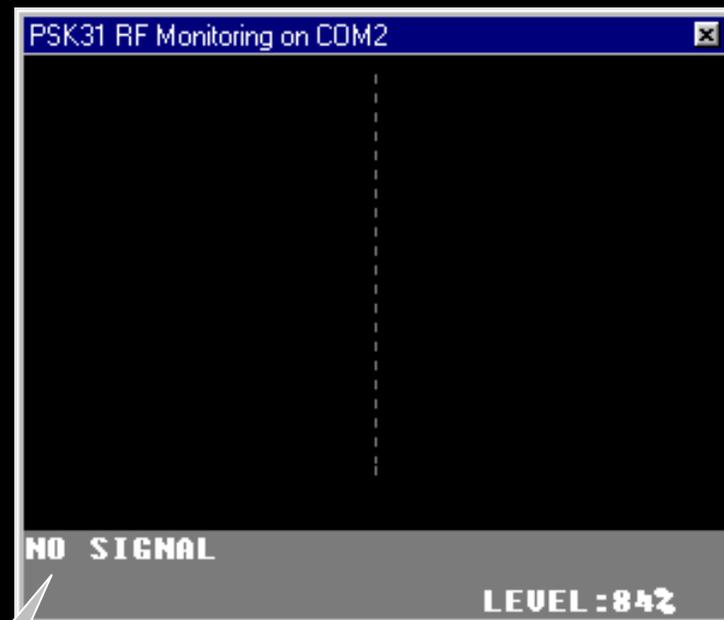
# A look at the PSKMeter Software

- When PSKMeter is launched, it displays the COM port it uses to connect to the PSKMeter hardware.
- If the hardware fails to respond, a message is displayed.
  - ◆ Wrong COM Port
  - ◆ No serial cable
  - ◆ No PSKMeter hardware
  - ◆ PSKMeter hardware not powered up



# A look at the PSKMeter Software

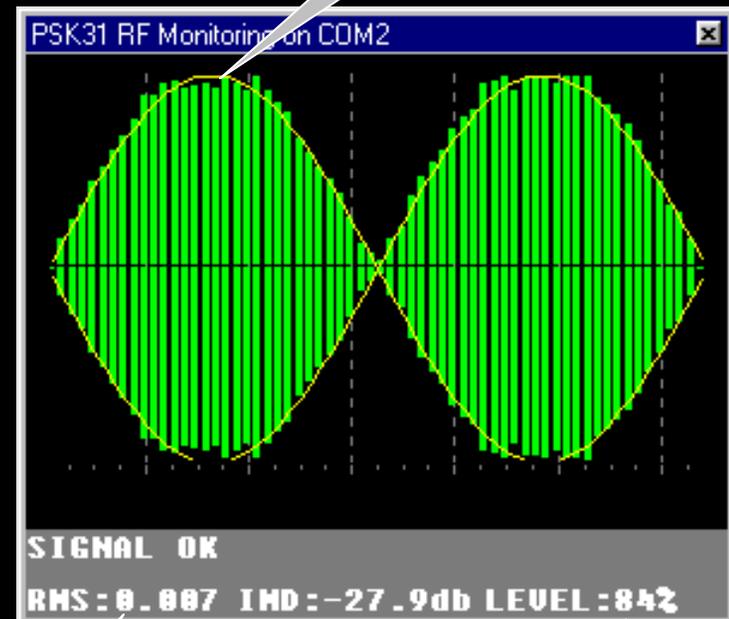
- If the hardware connects, the software periodically issues a request for waveform sample.
- When you are not in transmit mode, no signal is returned.



Absence of  
Signal

# A look at the PSKMeter Software

- In transmit mode, the RF output is periodically sampled and displayed.
- The best-fit sine wave curve is displayed. Variation of the signal from this sine wave is a measure of distortion.
- RMS variation from the sine wave is displayed.
- IMD is displayed.
- If RMS is small and IMD is large, the indication "SIGNAL OK" is displayed. Your signal is clean, and no adjustment (automatic or manual) is necessary.



Best fit  
sine wave

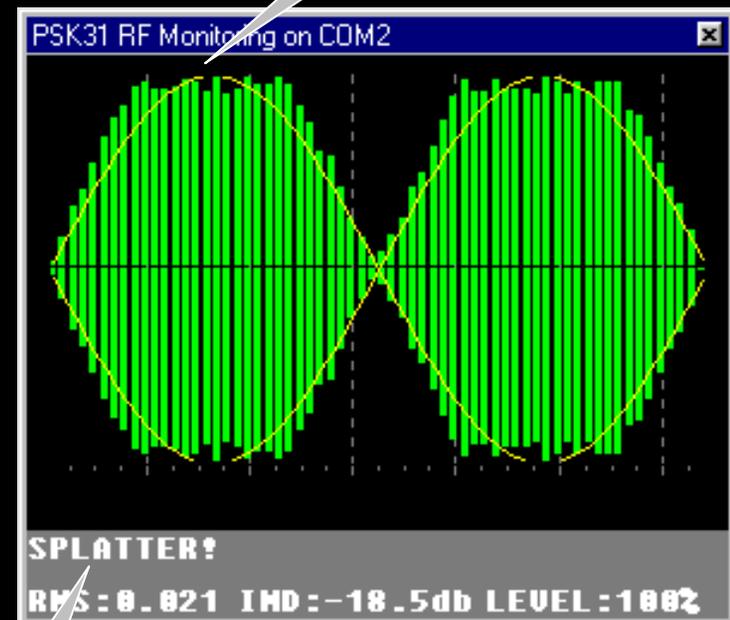
RMS  
variation of  
signal

Distortion  
figure

Sound  
Card Audio  
Level

# A look at the PSKMeter Software

- If you are over-driving your transmitter, your signal is going to show visible evidence of distortion.
- Example: flat-topping (clipping).
- The RMS will be large and the IMD will be small.
- The result is splatter, as indicated. You are causing interference to other portions of the band, and your signal will not be as copiable!
- When in automatic mode, the software will automatically reduce the sound card's audio level to completely remove the splatter.



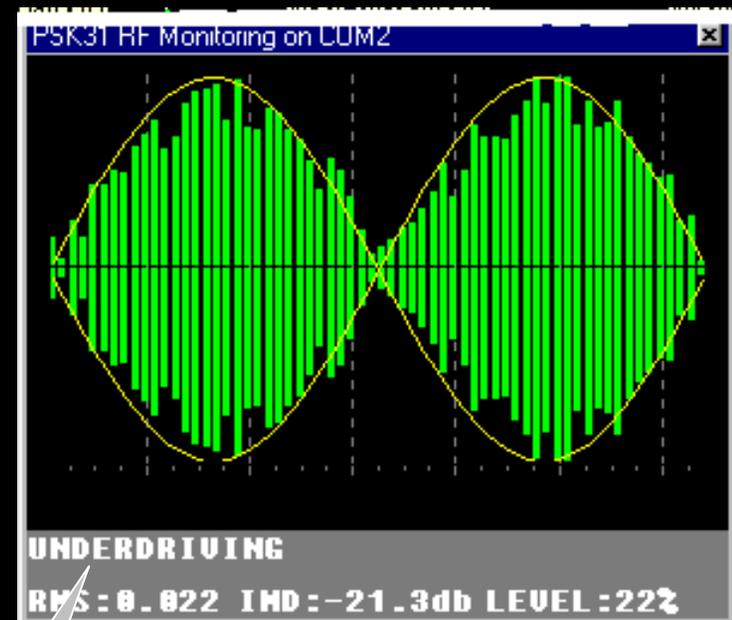
Signal shows "flattopping"

A warning is displayed

Too much audio!

# A look at the PSKMeter Software

- If you are under-driving your transmitter, your signal will be weak and noisy, but not distorted.
- Because your signal is weak, you will not be as copiable.
- When in automatic mode, the software will automatically increase the sound card's audio level, boosting your output power without introducing distortion.



A warning is displayed

Feeble audio

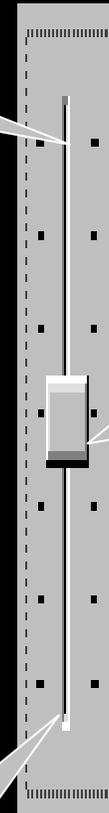
# PSKMeter Benefit:

- Bottom line: PSKMeter *automatically* finds the “sweet spot” in the audio level setting that outputs the highest power from your transmitter without creating splatter.

High Audio=  
Poor IMD (splatter),  
Poor copy

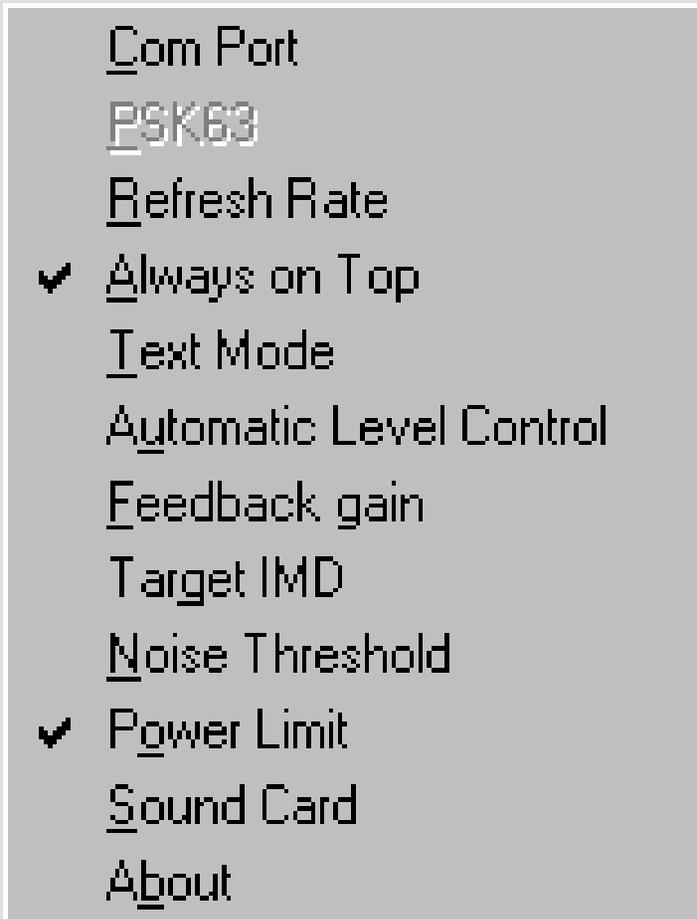
“Sweet spot”=  
Highest possible  
output power with  
good IMD (no  
splatter)

Low audio=  
Poor S/N (noisy),  
Poor copy



# PSKMeter User Interface

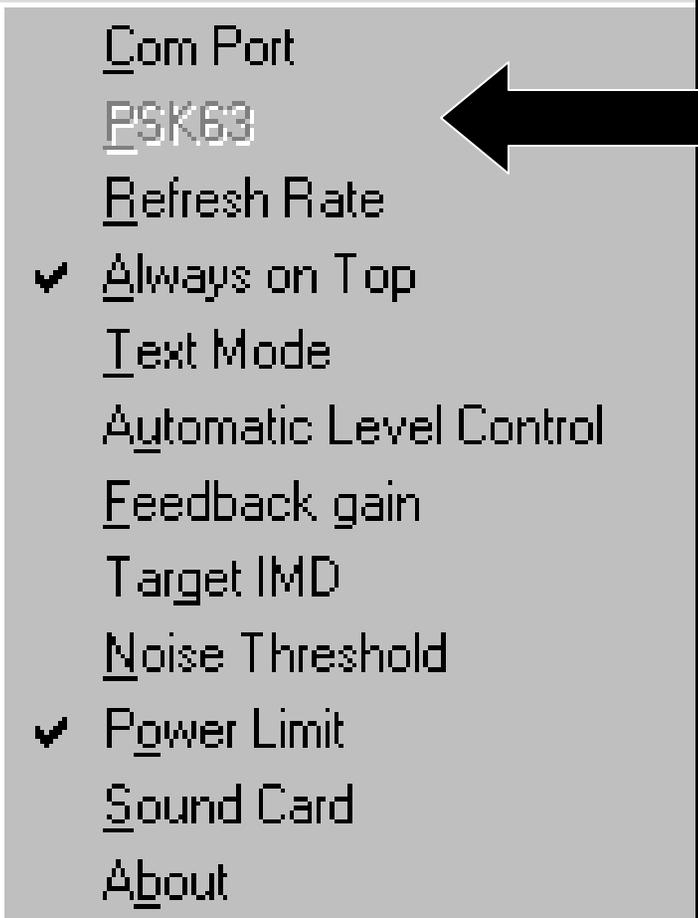
- Right-click on the meter brings up the user interface menu:
- Com port selection
- Mode selection (PSK31/PSK63)
- Refresh (sampling) rate
- Force window to be always visible
- View data numerically
- Force automatic control of audio level
- Gain selection
- Desired IMD selection
- Noise (floor) threshold
- Maximum power emission selection
- Sound card selection

A screenshot of the PSKMeter user interface menu, displayed in a light gray box with a thin white border. The menu items are listed vertically, with the first letter of each item underlined. Two items, 'Always on Top' and 'Power Limit', are preceded by a checkmark symbol. The items are: Com Port, PSK63, Refresh Rate, Always on Top, Text Mode, Automatic Level Control, Feedback gain, Target IMD, Noise Threshold, Power Limit, Sound Card, and About.

Com Port  
PSK63  
Refresh Rate  
✓ Always on Top  
Text Mode  
Automatic Level Control  
Feedback gain  
Target IMD  
Noise Threshold  
✓ Power Limit  
Sound Card  
About

# Mode Selection

- “Standard” mode is PSK31 (31.25 Hz = 32 msec per bit)
- Recently we have the ability to transmit PSK63 (62.50 Hz = 16 msec per bit)
- By selecting PSK63, a command is sent to the firmware that instructs it to use  $\frac{1}{2}$  msec sampling time instead of 1 msec. Everything else works the same way.



Com Port  
PSK63  
Refresh Rate  
✓ Always on Top  
Text Mode  
Automatic Level Control  
Feedback gain  
Target IMD  
Noise Threshold  
✓ Power Limit  
Sound Card  
About

# Maximum Power Selection

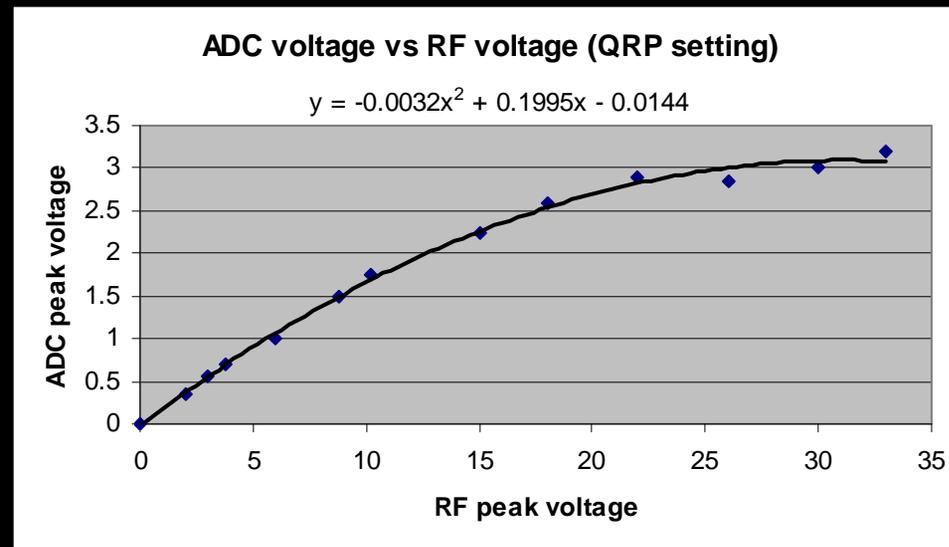
- PSKMeter is capable of controlling the power output to a specified level
- From  $P = E^2/2R = E^2/100$ , we know the peak RF voltage given an average power level (i.e.,  $E = 10 \sqrt{P}$ ).
- But what is the relationship between the peak voltage seen at the ADC to the peak RF voltage?

Com Port  
PSK63  
Refresh Rate  
✓ Always on Top  
Text Mode  
Automatic Level Control  
Feedback gain  
Target IMD  
Noise Threshold  
✓ Power Limit  
Sound Card  
About



# ADC voltage vs RF voltage

- Shown below is a measurement of the signal at the ADC as a function of the RF peak voltage:



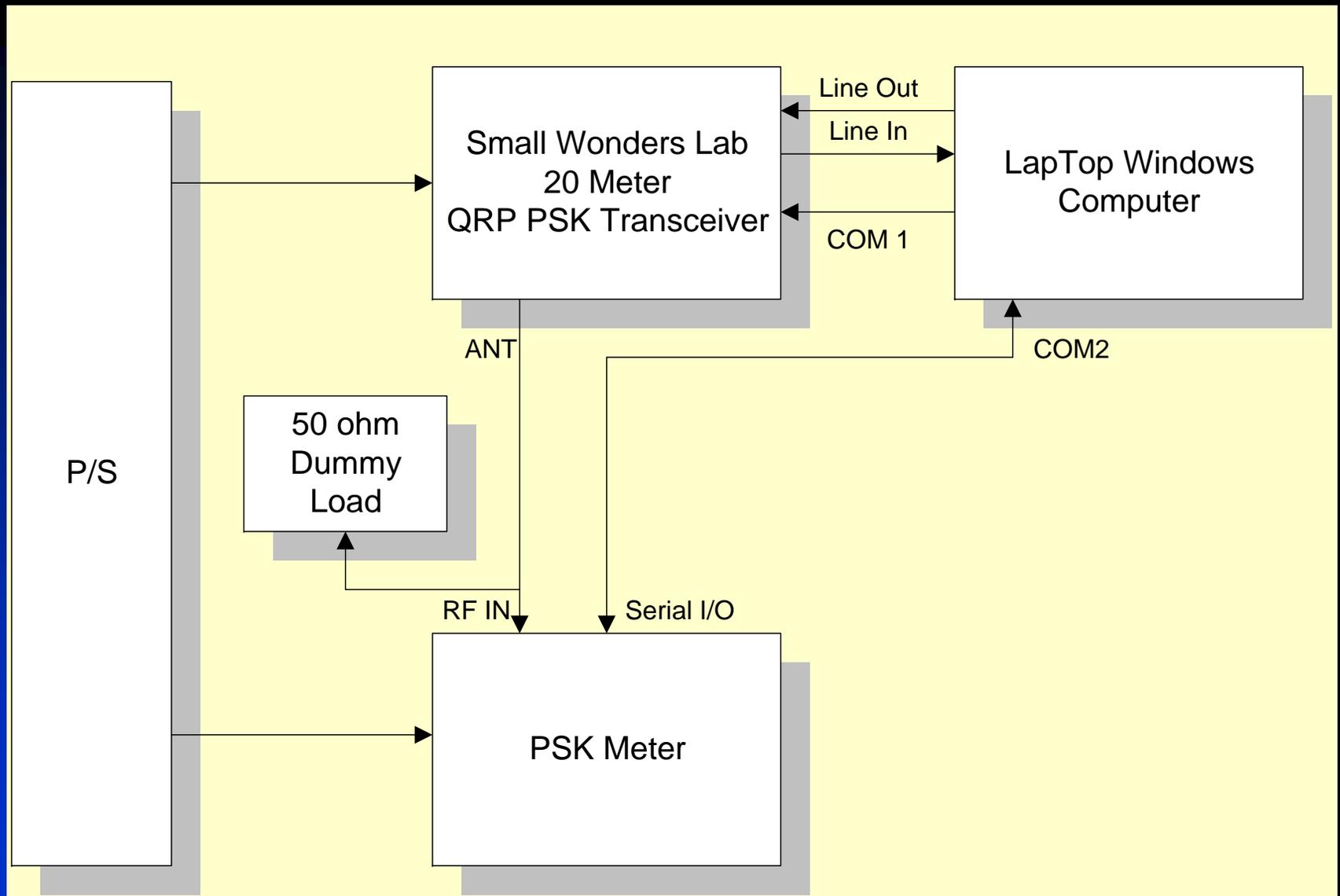
- Notice that the relationship is linear only for low power. The function is well approximated by a cubic polynomial over a wide range of power.

# Maximum Power Selection

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- Thus, given a desired power level, the software computes the RF peak voltage, then using this polynomial fit to the empirical data determines the voltage at the input to the ADC.
- The software then sets the audio level to obtain this voltage, thereby achieving the desired radiated power.

# Demonstration Time!



# Building one for yourself

- Circuit diagram, parts list, assembly instructions, pskmeter.exe and firmware hex file are available at no charge at <http://www.ssiserver.com/info/pskmeter>.
- Alternatively, PSKMeter is available in kit form; ordering can be done via a link at the same web site page shown above (or, I have some kits with me today).
- Questions and comments can be sent to [george@softsci.com](mailto:george@softsci.com).



# Future Enhancements

- Use one of the RS232 control lines to power the circuit—eliminates the need of a power adapter (wall wart)(adds dependencies on specific com port implementations—will it work for all computers and laptops?)
- Use an op-amp IC with diodes to obtain precision rectification—increases accuracy of the IMD calculation (714 lacks bandwidth, LM318 is fussy and tends to oscillate; high speed op-amp will add to cost)

# Future Enhancements

- Port the existing pskmeter.exe application to a DLL (with a published API), to be accessed at first by Digipan and MixW and would allow Digipan and MixW to embed and incorporate PSKMeter functionality.
- Modify hardware to pass the PTT control signal to the rig interface--only a single COM port would be required. PSKMeter would be “in line” between the computer and the rig interface. Also, PSKMeter would know when the transmitter is keyed up eliminating the need to poll the hardware when in receive mode.

# Future Enhancements

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- Include rig interface circuitry on the PSKMeter printed circuit board, further reducing the complexity to the user (and the need for a separate rig interface) (adds parts and costs, but would be optional)

# Where to go for more info

[www.ssiserver.com/info/pskmeter](http://www.ssiserver.com/info/pskmeter). Download a copy of this presentation in PDF format (starting Monday), assembly instructions, circuit diagrams, firmware and software, software upgrades, order PSKMeter kits, read FAQs, etc.

Attend Steve Ford's WB8IMY talk "Introduction to PSK31" tomorrow at 1pm in Windor I.

*Wireless Digital Communications: Design and Theory*, Tom McDermott, N5EG

[www.arrl.org](http://www.arrl.org) (members only, search for PSK31)

[www.psk31.com](http://www.psk31.com)

See you on 14.070 PSK-31!

You've been a great audience!

73's from  
George, KF6VSG